



MPEG-1 lag 1, 2 og lag 3

Sverre Holm

Basert på presentasjon laget av
Torbjörn Ekman, 2005 (nå på NTNU)

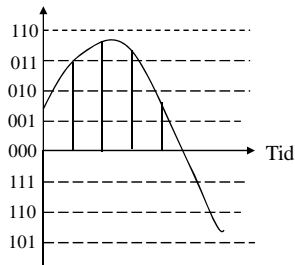


MPEG audiokoding

- Motivasjon for de fleste kapitlene i
Ambardar, Digital signal processing: A
Modern Introduction, Thomson, 2007.



Digital representation of Sounds Pulse Coded Modulation (PCM)

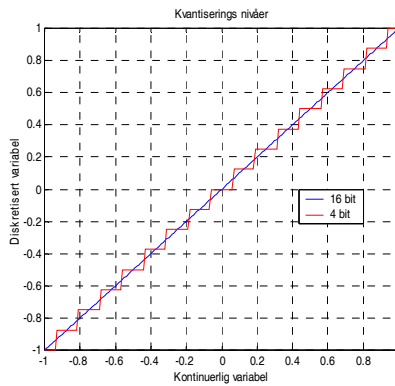


16 bit kvantisering gir


$$2 \cdot 2 \cdot 2 \cdot \dots \cdot 2 = 2^{16} = 65\,536 \text{ nivåer}$$


Ved 44100 samples per sek, blir bitraten:

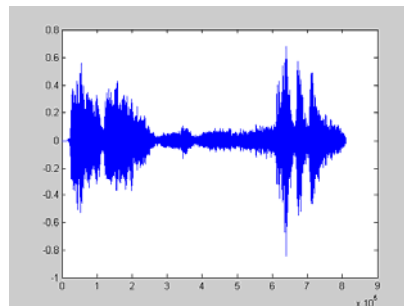
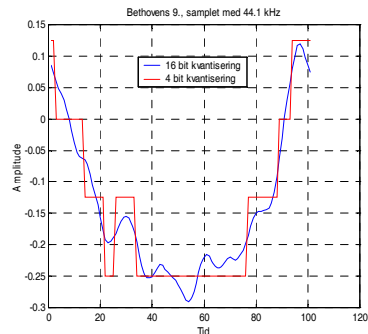
$$16 \cdot 44100 = 705\,600 \text{ bits/s} = \text{halv CD-rate}$$



Beethovens 5. symfoni

16 bit kvantisering 
 $2^{16} = 65536$ nivåer

4 bits kvantisering 
 $2^4 = 16$ nivåer





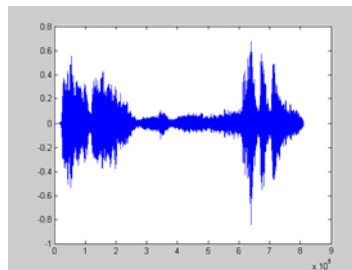
Hvorfor høres det så ille ut?

- Problem: Bare noen få kvantiseringsnivåer => for stor avrundingsfeil. Ofte at signalet settes til null da nivået var lavere enn laveste kvantiseringsnivå
- Kvantisering og sampling:
Kap 7: Digital behandling av analoge signaler
- Mulig løsning: Skaler blokker av data slik at maximumsverdien alltid utnytter hele dynamikkområdet til kvantisereren
- Kostnad: Må sende over skalafaktorer
 - NICAM, Near Instantaneous Companded Audio Multiplex: format for digital lyd over analog TV.
 - Blokk lengde 32 samples, 3 bit pr blokk sideinfo. Stereo kodes med 10 av 14 bit ved samplingsrate 32 kHz => 728 kbit/s.
 - Variant av adaptiv differensiell puls kode modulasjon



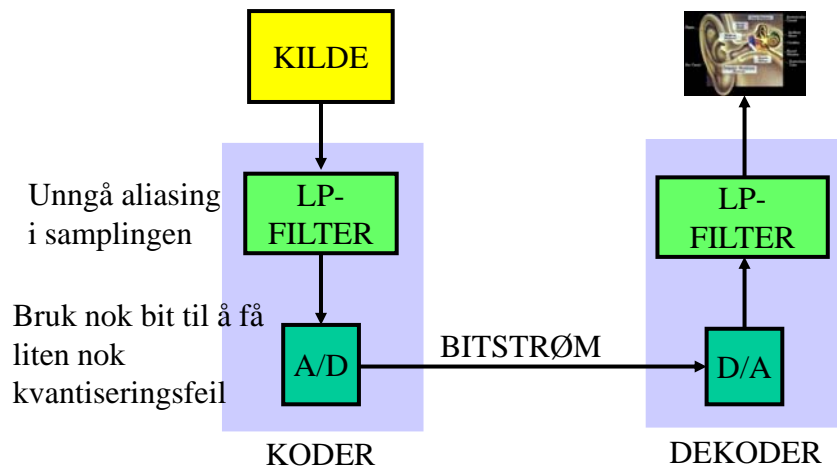
Stasjonærhet - tidsinvarians

- = Egenskaper varierer ikke med tiden, **kap 3: Tids-domene analyse**
- Forutsettes i det meste av analyser
- Tale er korttids stasjonær, dvs bare over ca 20 ms,
 - Endres $1/20 \times 10^{-3} = 50$ ganger pr sekund





Direktesampling (PCM)



Unngå aliasing
i samplingen

Bruk nok bit til å få
liten nok
kvantiseringsfeil



Bitrater

- CD: $44.1 * 2 * 16 = 1.411$ Mbit/s
– 4 bit: 25% => 350 kbit/s låter forferdelig
- MP3, AAC etc: 128 kbit/s ~ CD/12
- Hva er det lure trikset?



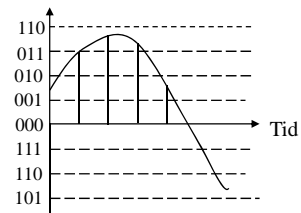
MPEG-1 Audio

Psychoacoustics in sound compression

- How do we hear?
- Digital representation of sounds
- Sound compression
- Psychoacoustics
 - Masking
 - Adaptive quantization
 - Bit allocation
- Filterbanks



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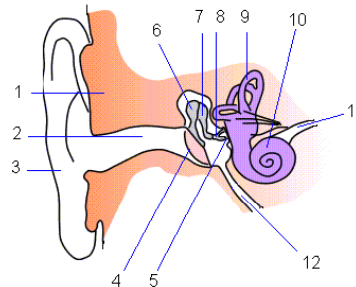


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Øret

1. Tinning
2. Øregang
3. Ytre øre
4. Trommehinne
5. Ovale vindu
6. Hammeren
7. Ambolt
8. Stigbøyle
9. Bueganger
10. Sneglehuset
11. Hørselnerve
12. Øretrompeten



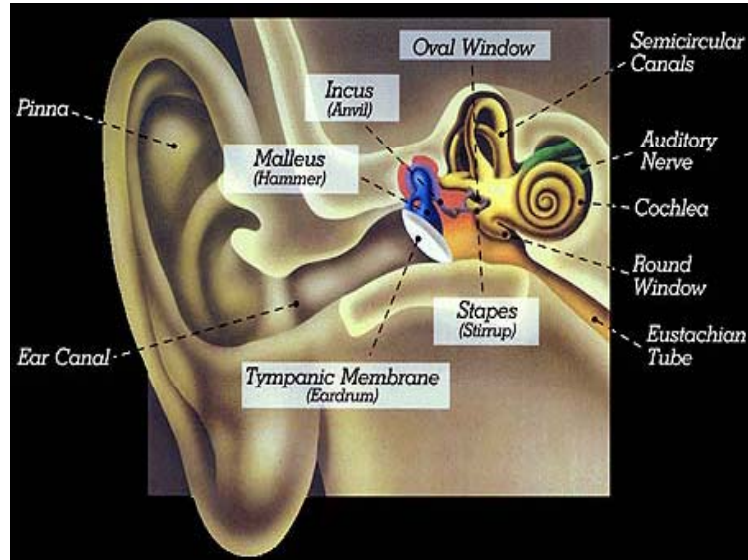
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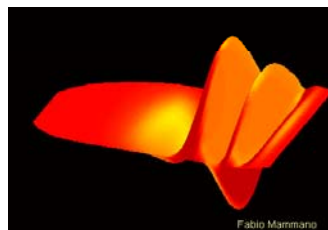
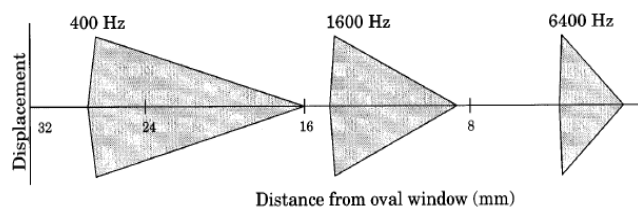


The Ear



The frequency filters of the ear: Mapping frequency to a location

Unwound
cochlea

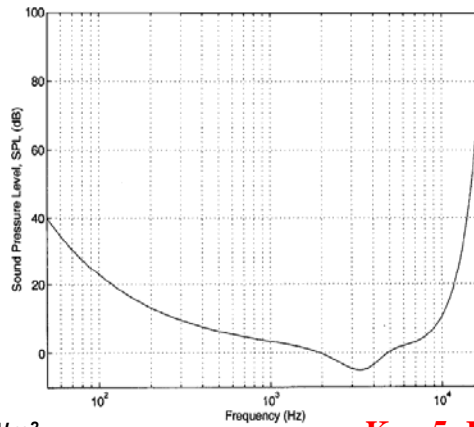


Fabio Marriana

Kap 5: Frekvensanalyse



Threshold for audible sounds

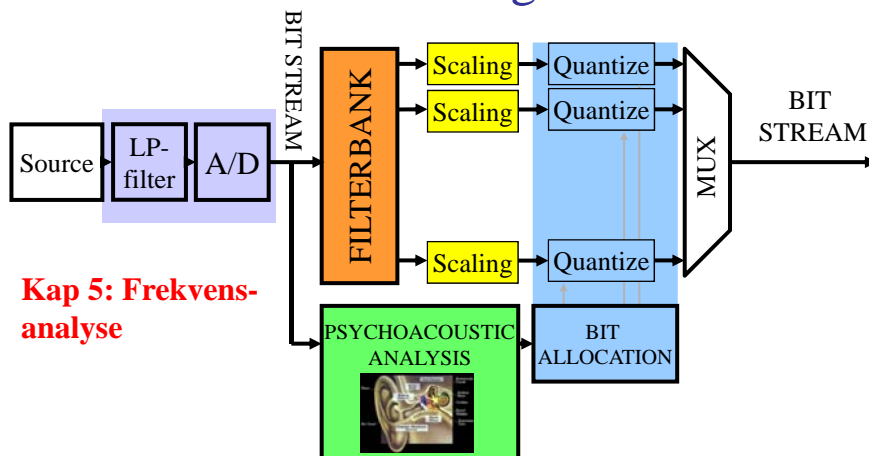


Reference 0 dB:
 $20 \mu\text{Pa} = 2 \cdot 10^{-5} \text{ N/m}^2$

Kap 5: Frekvensanalyse



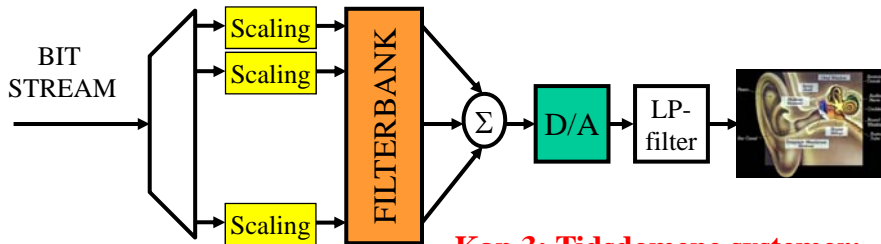
Filterbank Approach Encoding



**Kap 5: Frekvens-
analyse**



Decoding is much simpler

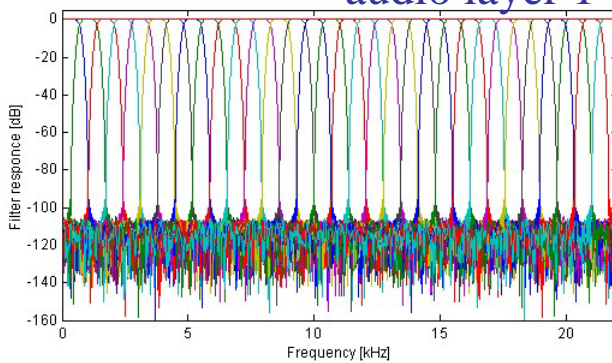


**Kap 3: Tidsdomene systemer:
linearitet**

Kap 3: Inverse systemer



Filterbanks in MPEG-1 audio layer 1-3



**Kap 5: Frekvens-
analyse av systemer**

Kap 6: Digitale filtre

**Kap 10: FIR
Filterdesign**

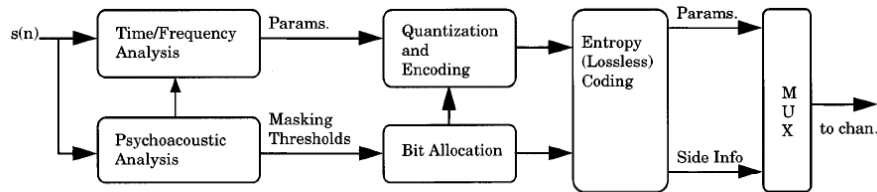
- Polyphase filterbank
- 32 subbands, e.g. bw
 $44100/2/32 = 689$ Hz
- 512 tap FIR-filters
- 80 adds and mults per output

- Equal width
- Not perfect reconstruction
- Frequency overlap

Kap 4: z-transform



What is this Psychoacoustics that is used in the Encoder ?



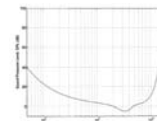
Kap 8: Diskret Fourier Transform; Estimering av effektspektrum



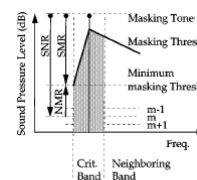
Masking

We do not hear all sounds.

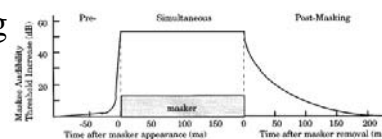
1. Absolute threshold of hearing: →
2. Masking: One sound is inaudible in the presence of another sound.



1. Simultaneous masking
 - Noise Masking Tone
 - Tone Masking Noise
 - Noise Masking Noise



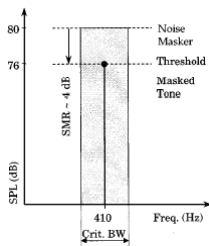
2. Nonsimultaneous masking
 - Pre masking (2 ms)
 - Post masking (100 ms)





Noise Masking Tone

| | | | | |
|---|------------------------------------|------------------------------------|----------------------|----------------------|
| Filtered Noise Center 410 Hz Width 111 Hz | Tone 1, 820 Hz 5 dB below noise | Tone 2, 410 Hz 5 dB below noise | Noise + Tone 1 | Noise + Tone 2 |
| | | | Not masked | Masked |



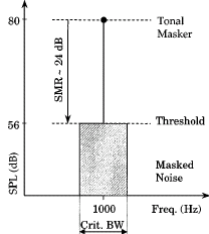
You can not hear a sinusoid that lies in the same critical band as a filtered noise if the sound pressure level is below a certain threshold.

This effect also stretches out beyond the critical band.



Tone Masking Noise

| | | | | |
|---|---------------|---------------|----------------------|----------------------|
| Filtered Noise Center 1 kHz Width 162 Hz 15 dB below | Tone 1, 2 kHz | Tone 2, 1 kHz | Noise + Tone 1 | Noise + Tone 2 |
| | | | Not masked | Masked |



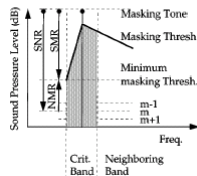
You can not hear a filtered noise that lies in the same critical band as a sinusoid if the sound pressure level is below a certain threshold.

This effect also stretches out beyond the critical band.



Exploit Masking

- If a sound is masked we can't hear it.



- Make a frequency analysis of the signal and find the masking threshold.
- Put the quantization noise under the masking threshold and we won't hear the quantization.

Kap 8: DFT, Fast Fourier transform, Estimering av effektspektrum

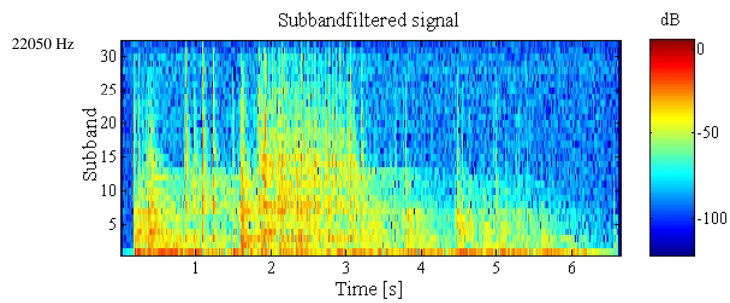
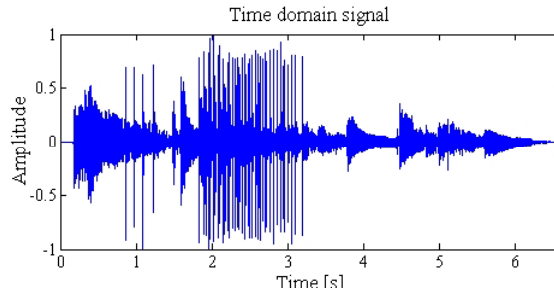


Bit Allocation and Masking

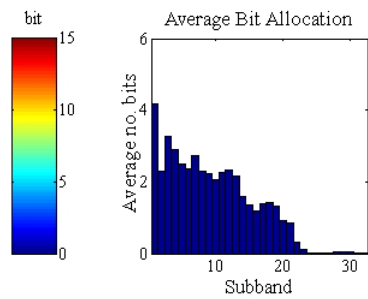
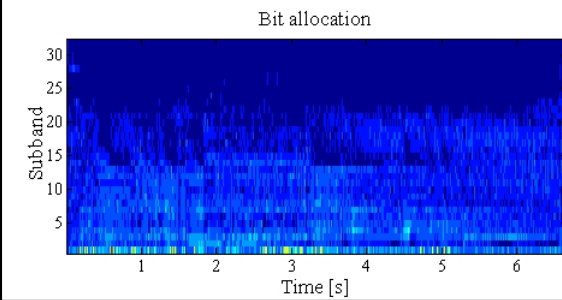
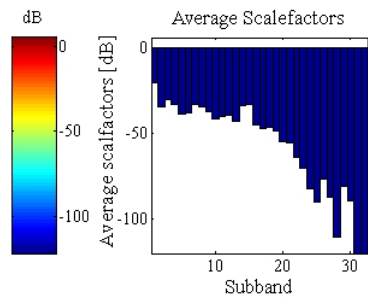
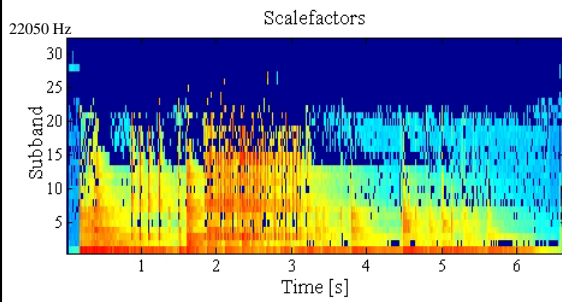
- The masking threshold in each subband gives the Just Noticeable Distortion (JND) limit for that band.
- Bits are assigned to subbands so that the quantization noise falls below or as little over the JND as possible.
- Then the Signal to Quantization noise Ratio (SQR) falls below JND



Castanets and Guitar

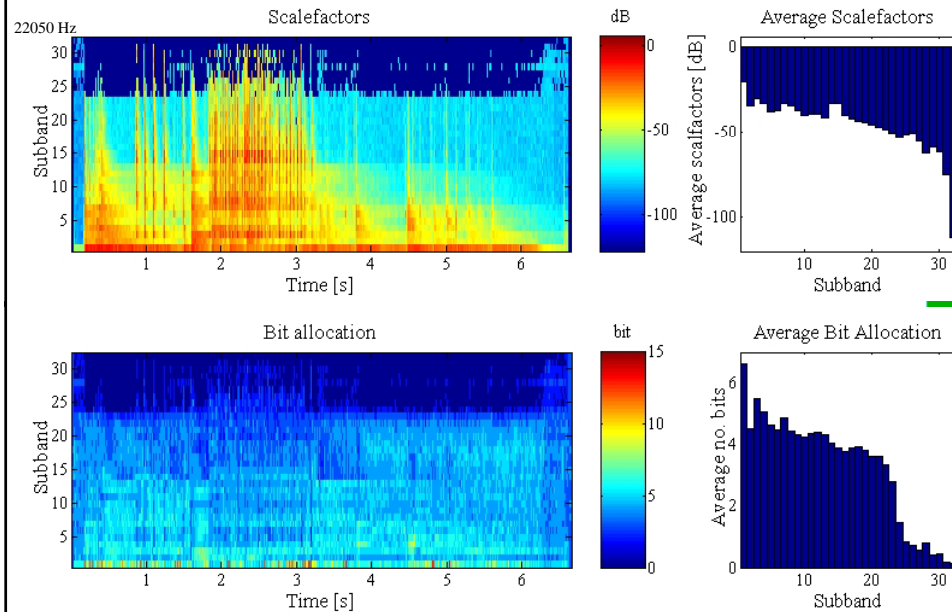


Bit allocation with 2 bits per sample





Bit allocation with 4 bits per sample



Kapitler i læreboka – MP3-koding

| | | |
|----|--|---|
| 1 | Overview | <ul style="list-style-type: none"> Praktiske eksempler som mp1/mp3 |
| 2 | Discrete Signals | <ul style="list-style-type: none"> Musikk og talesignal |
| 3 | Time-Domain Analysis | <ul style="list-style-type: none"> Linearitet: kan dele i frekvensbånd, behandle for seg og så sette sammen igjen Tidsinvarians gjelder bare over kort tid for musikk og tale. Differanseligninger: FIR filtre |
| 4 | z-Transform Analysis | <ul style="list-style-type: none"> Analyse av filter i filterbanken: nøkkel til å få til filterdesign |
| 5 | Frequency Domain Analysis | <ul style="list-style-type: none"> Frekvensdomene er sentralt i modell av hørsel Frekvensselektive filtre: båndpassfiltre Inverse systemer: kan dele i bånd i koder og addere sammen igjen i dekker |
| 6 | Filter Concepts | <ul style="list-style-type: none"> Filterstrukturer, hvordan implementere filterbank i koder og dekker |
| 7 | Digital Processing of Analog Signals | <ul style="list-style-type: none"> A/D-analyse: kvantiseringsstøy ved direkte sampling Multirate system: Hvert delfilter kan nedsamples pga bare 1/32 av total båndbredde => trenger bare 1/32 samplerate per filter |
| 8 | The Discrete Fourier Transform and Its Applications. | <ul style="list-style-type: none"> Frekvensanalyse av signaler FFT brukes i estimering av spektrum i koder. Må estimere korttidspektrum for å gjøre adaptiv bittildeling |
| 9 | Design of IIR Filters. | |
| 10 | Design of FIR Filters. | <ul style="list-style-type: none"> Hvordan finne koeffisienter til bp-filtrene i filterbanken? |
| 11 | MATLAB Examples | |
| A | Useful Concepts from Analog Theory | |