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# MPEG-1 lag 1, 2 og lag 3

Sverre Holm

Basert på presentasjon laget av  
Torbjörn Ekman, 2005 (nå på NTNU)



# MPEG audiokoding

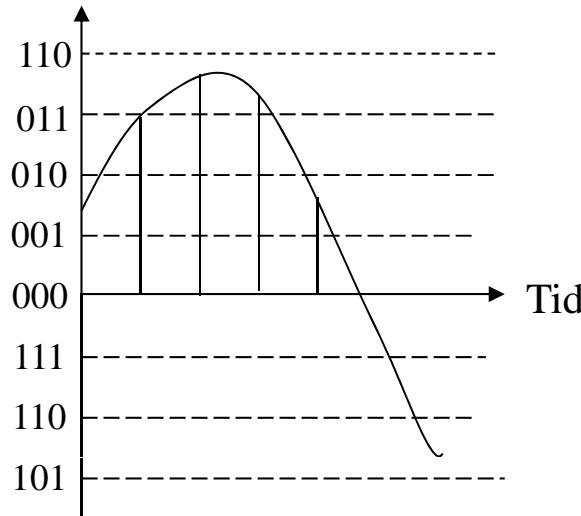
- Motivasjon for de fleste kapitlene i Ambardar, Digital signal processing: A Modern Introduction, Thomson, 2007.





# Digital representation of Sounds

## Pulse Coded Modulation (PCM)

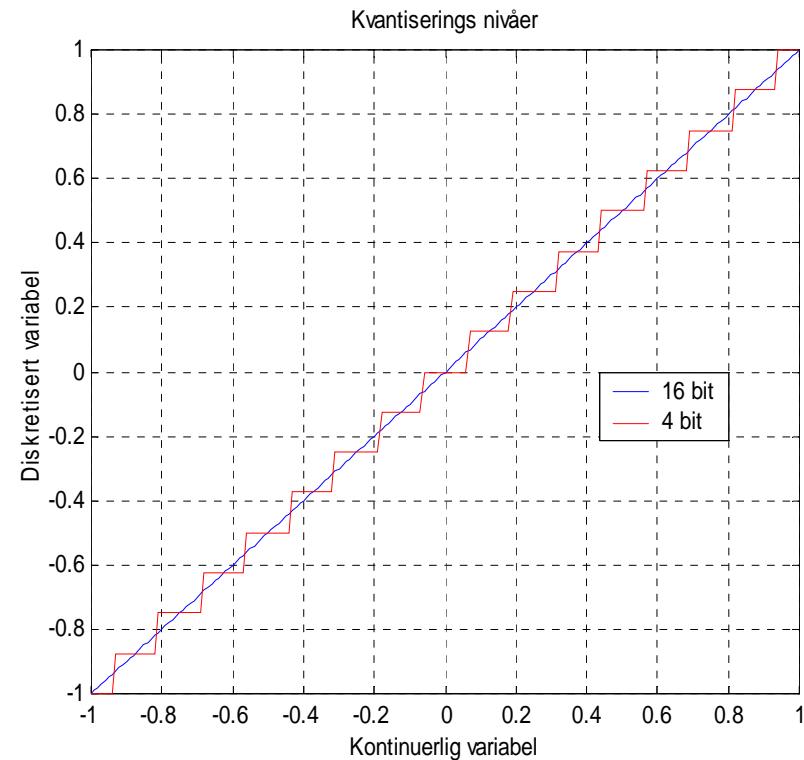


16 bit kvantisering gir

$$2 \cdot 2 \cdot 2 \cdot \dots \cdot 2 = 2^{16} = 65\,536 \text{ nivåer}$$

Ved 44100 samples per sek, blir bitraten:

$$16 \cdot 44100 = 705\,600 \text{ bits/s} = \text{halv CD-rate}$$



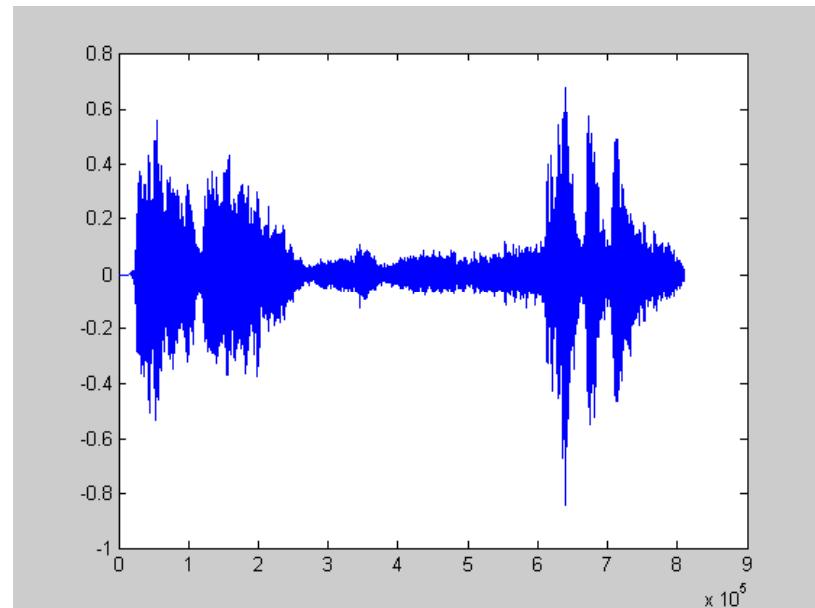
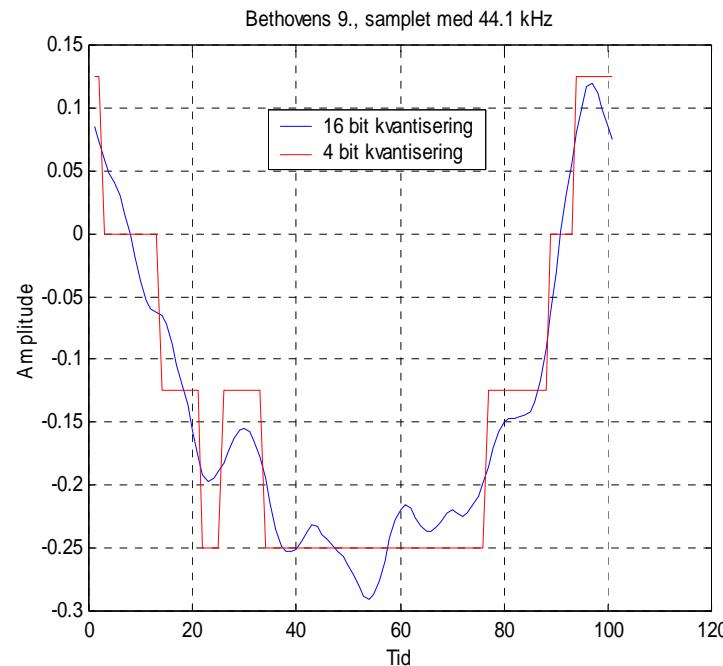


# Beethovens 5. symfoni

16 bit kvantisering  
 $2^{16} = 65536$  nivåer



4 bits kvantisering  
 $2^4 = 16$  nivåer





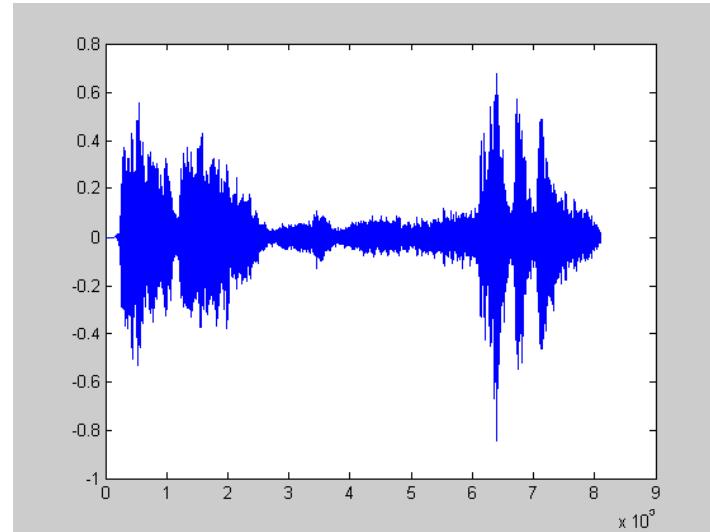
# Hvorfor høres det så ille ut?

- Problem: Bare noen få kvantiseringsnivåer => stor avrundingsfeil. Ofte at signalet settes til null da nivået var lavere enn laveste kvantiseringssnivå
- Kvantisering og sampling:  
**Kap 7: Digital behandling av analoge signaler**
- Mulig løsning: Skaler blokker av data slik at maximumsverdien alltid utnytter hele dynamikkområdet til kvantisereren
- Kostnad: Må sende over skalafaktorer
  - NICAM, Near Instantaneous Companded Audio Multiplex: format for digital lyd over analog TV.
  - Blokklengde 32 samples, 3 bit pr blokk sideinfo. Stereo kodes med 10 av 14 bit ved samplingsrate 32 kHz => 728 kbit/s.
  - Variant av adaptiv differensiell puls kode modulasjon



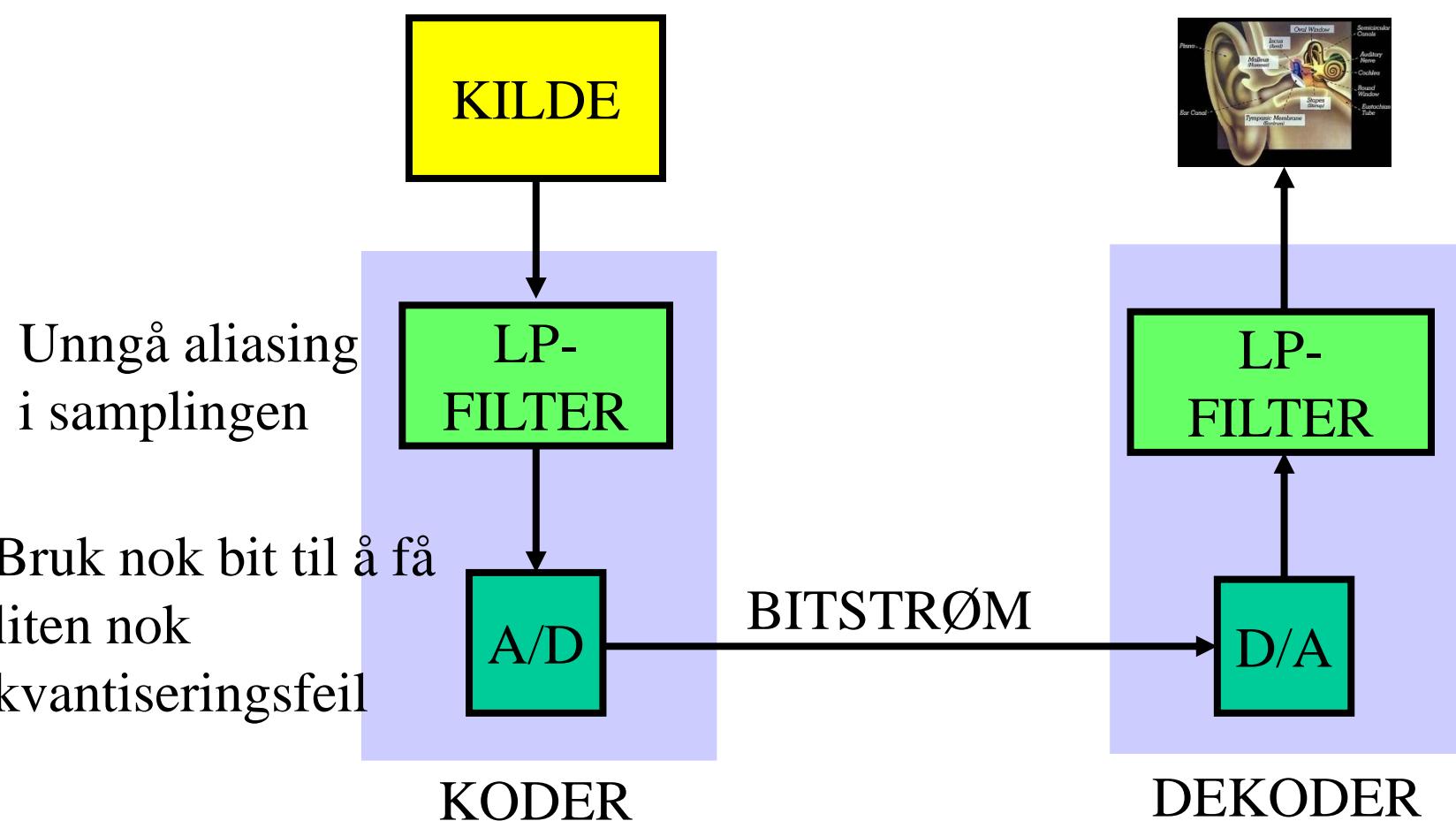
# Stasjonæritet - tidsinvarians

- = Egenskaper varierer ikke med tiden, **kap 3: Tids-domene analyse**
- Forutsettes i det meste av analyser
- Tale er korttids stasjonær, dvs bare over ca 20 ms,
  - Endres  $1/20e-3=50$  ganger pr sekund





# Direktesampling (PCM)





## Bitrater

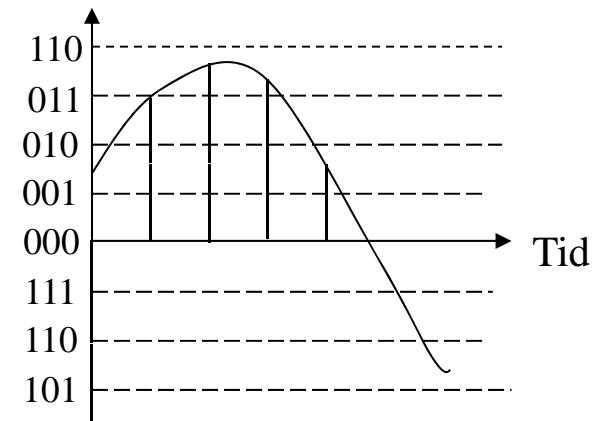
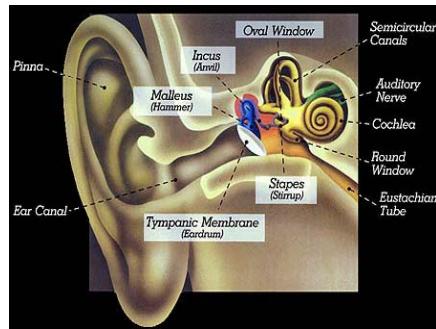
- CD:  $44.1 * 2 * 16 = 1.411 \text{ Mbit/s}$ 
  - 4 bit: 25%  $\Rightarrow 350 \text{ kbit/s}$  låter forferdelig
- MP3, AAC etc:  $128 \text{ kbit/s} \sim \text{CD}/12$
- Hva er det lure trikset?



# MPEG-1 Audio

## Psychoacoustics in sound compression

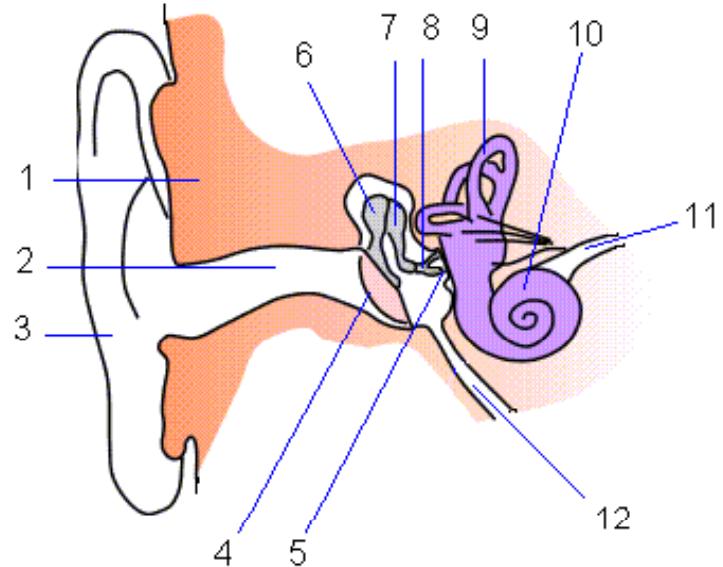
- How do we hear?
- Digital representation of sounds
- Sound compression
- Psychoacoustics
  - Masking
  - Adaptive quantization
  - Bit allocation
- Filterbanks





# Øret

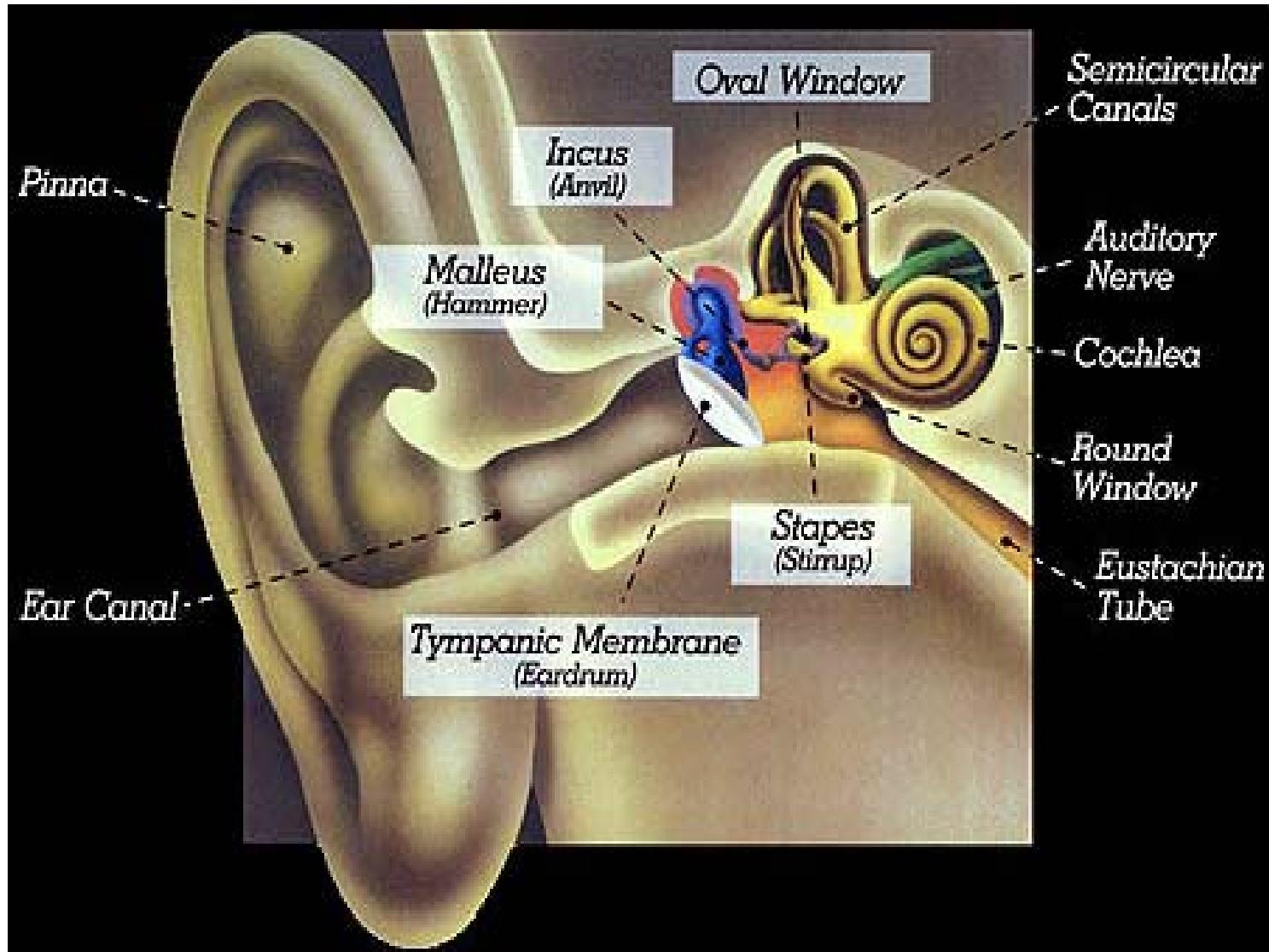
1. Tinning
2. Øregang
3. Ytre øre
4. Trommehinne
5. Ovale vindu
6. Hammeren
7. Ambolt
8. Stigbøylen
9. Bueganger
10. Sneglehuset
11. Hørselnerve
12. Øretrompeten



Wikimedia Commons



# The Ear

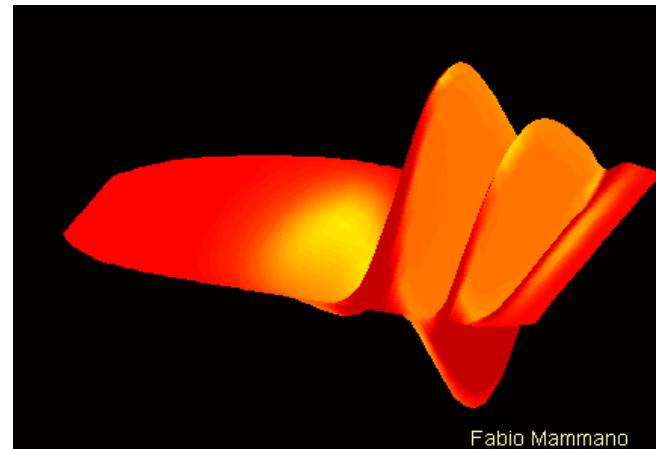
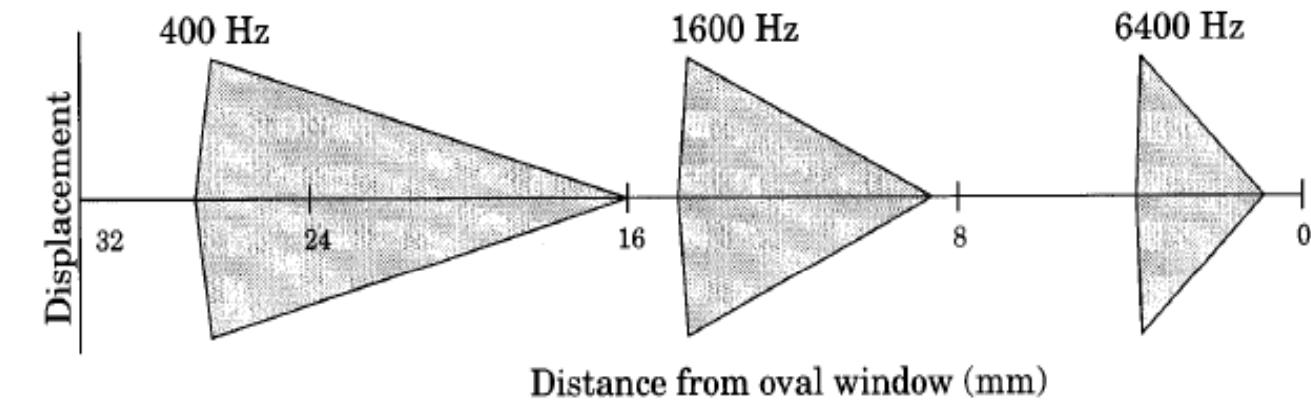




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# The frequency filters of the ear: Mapping frequency to a location

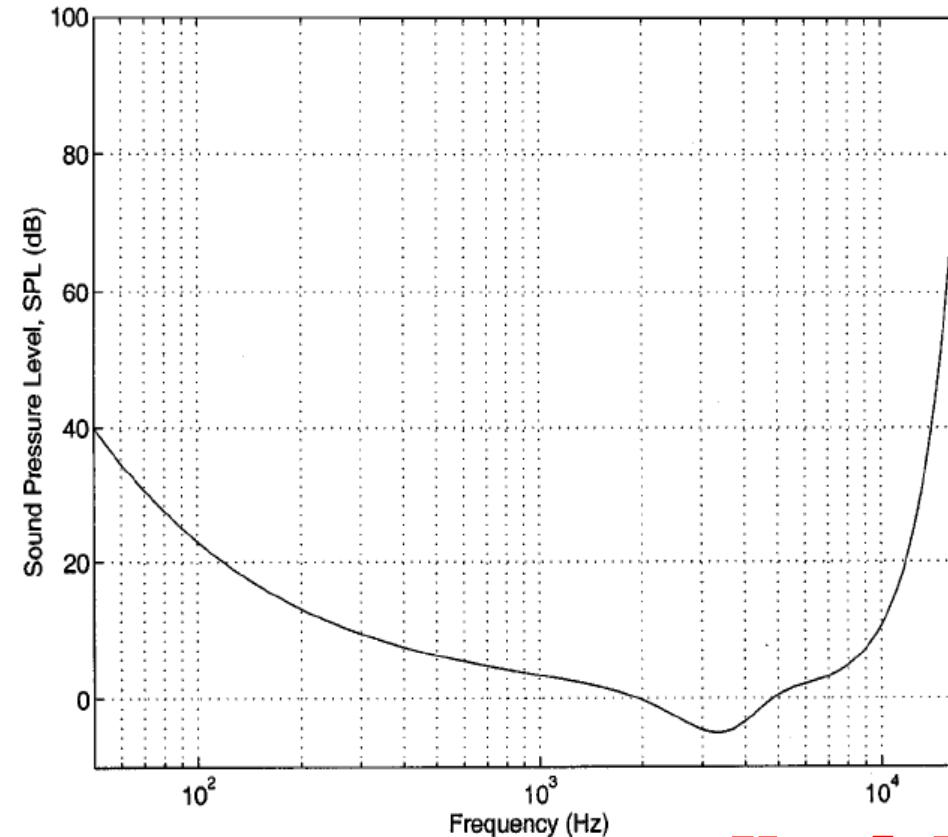
Unwound  
cochlea



**Kap 5: Frekvensanalyse**



# Threshold for audible sounds



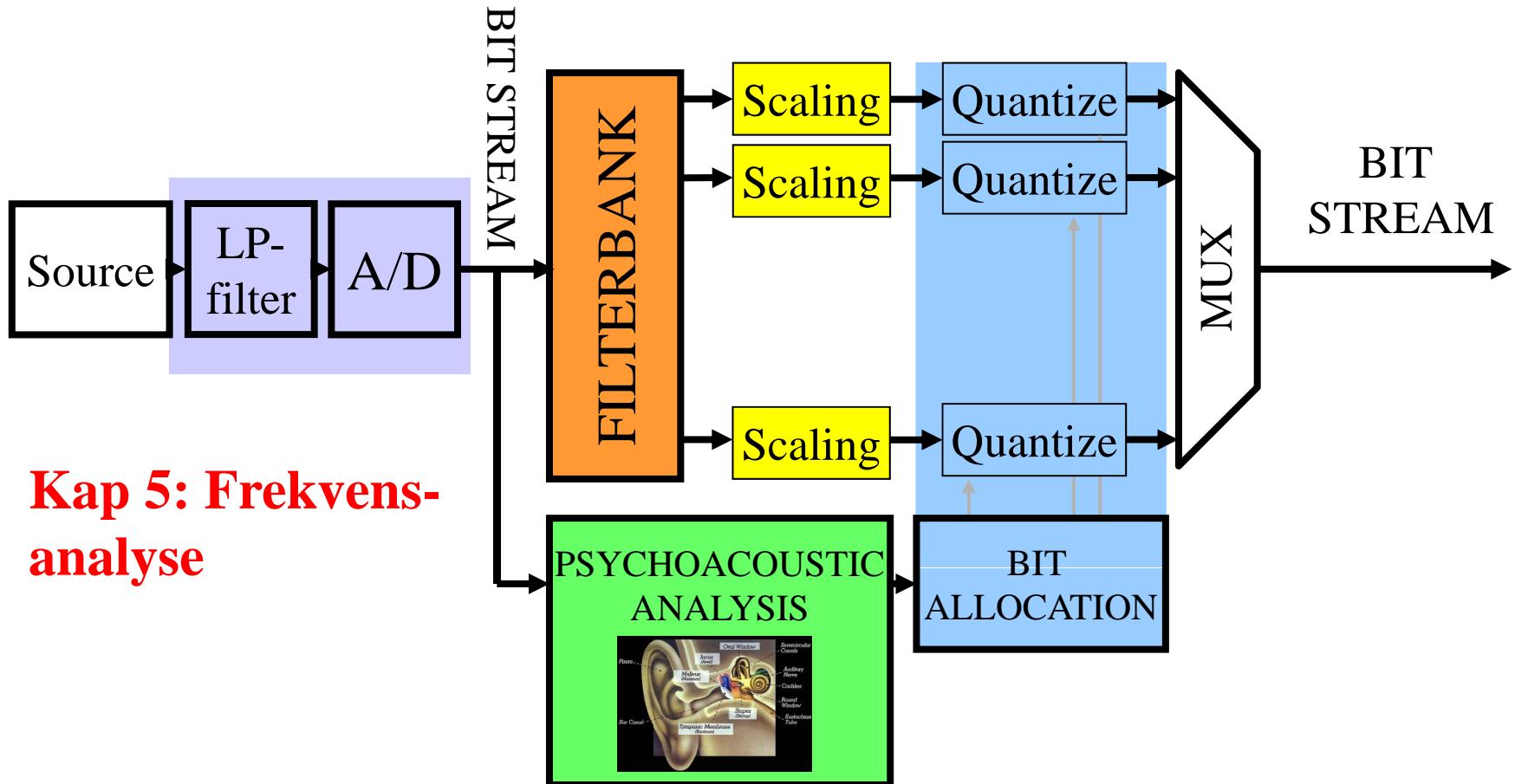
**Reference 0 dB:**  
 $20 \mu\text{Pa} = 2 \cdot 10^{-5} \text{ N/m}^2$

**Kap 5: Frekvensanalyse**



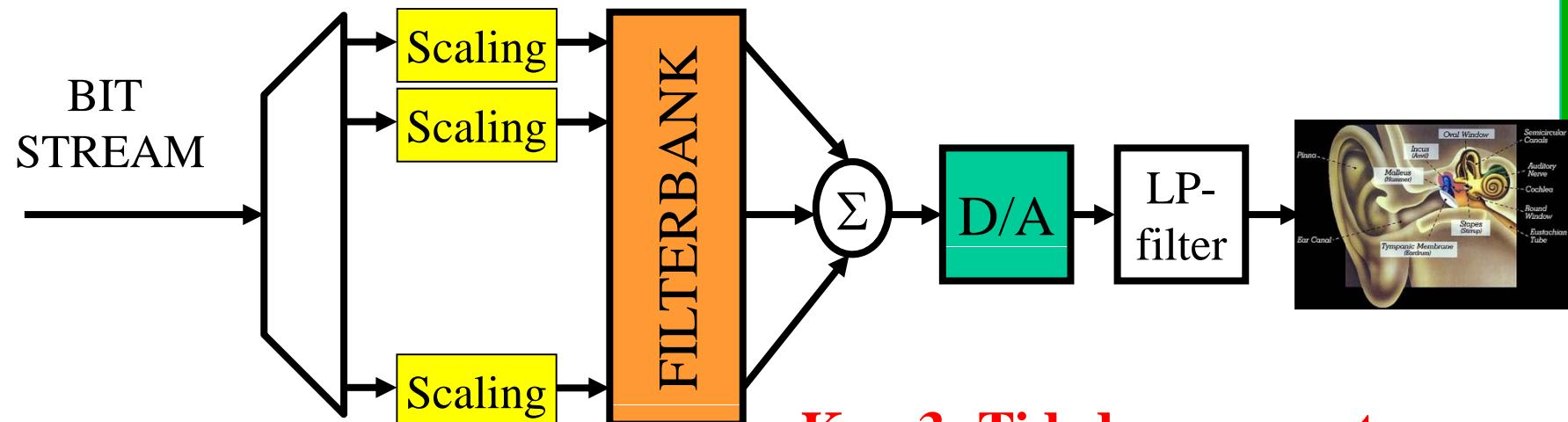
# Filterbank Approach

## Encoding





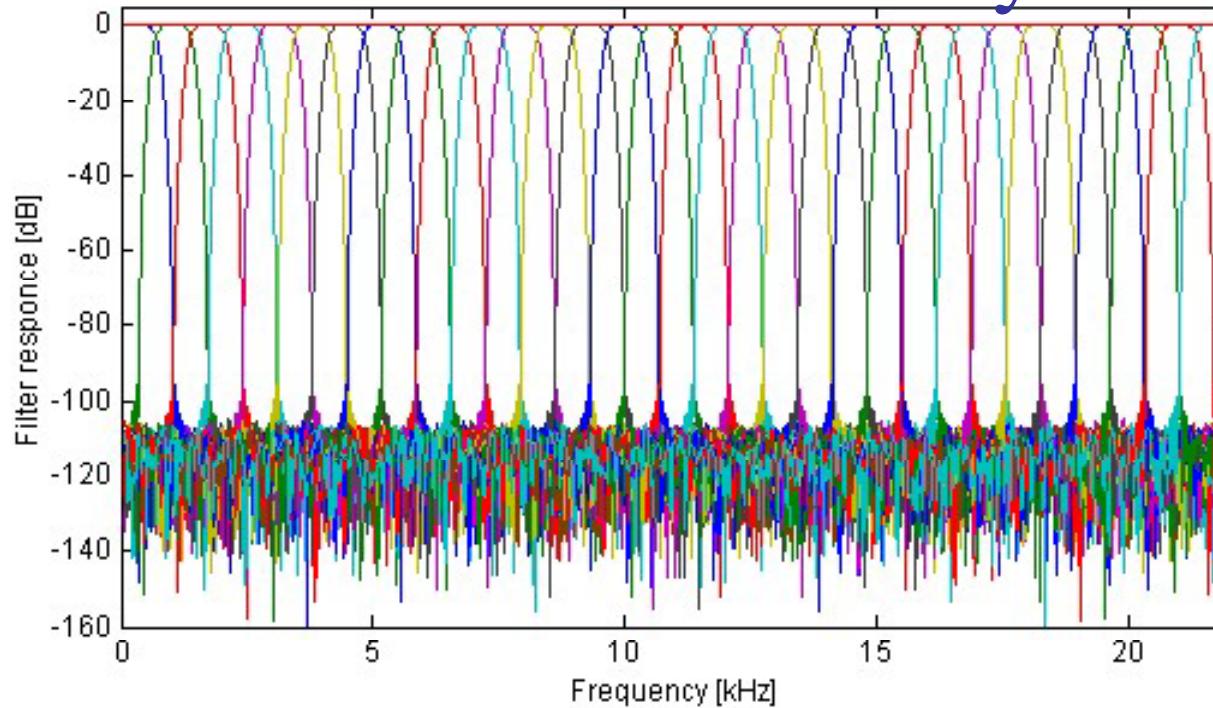
# Decoding is much simpler



**Kap 3: Tidsdomene systemer:  
linearitet**  
**Kap 3: Inverse systemer**



# Filterbanks in MPEG-1 audio layer 1-3



- Polyphase filterbank
- 32 subbands, e.g. bw  $44100/2/32 = 689$  Hz
- 512 tap FIR-filters
- 80 adds and mults per output

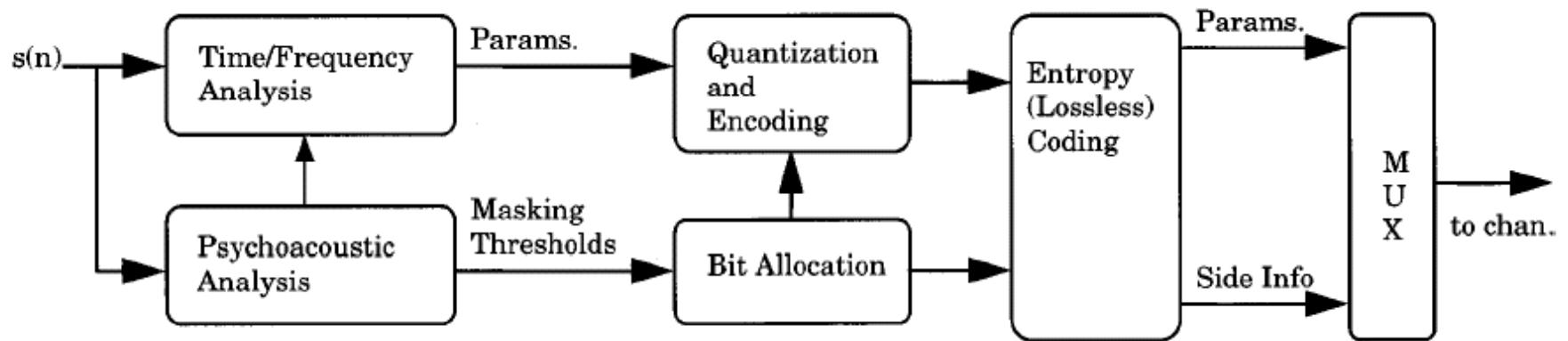
- Equal width
- Not perfect reconstruction
- Frequency overlap

**Kap 4: z-transform**

**Kap 5: Frekvens-analyse av systemer**  
**Kap 6: Digitale filtre**  
**Kap 10: FIR Filterdesign**



# What is this Psychoacoustics that is used in the Encoder ?



**Kap 8: Diskret Fourier Transform;  
Estimering av effektspektrum**



# Masking

We do not hear all sounds.

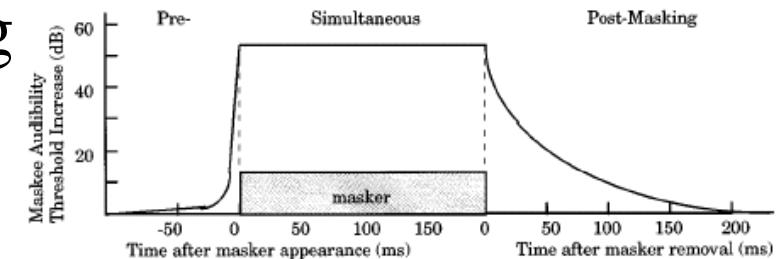
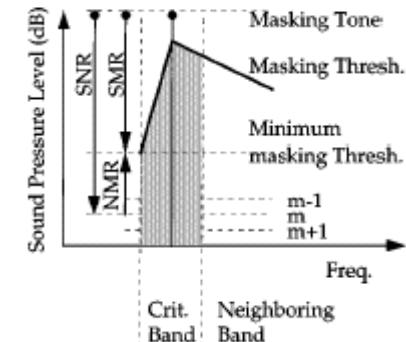
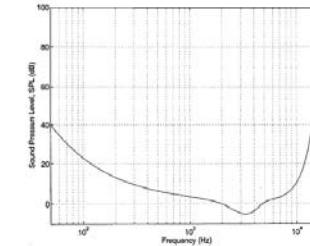
1. Absolute threshold of hearing: →
2. Masking: One sound is inaudible in the presence of another sound.

## 1. Simultaneous masking

- Noise Masking Tone
- Tone Masking Noise
- Noise Masking Noise

## 2. Nonsimultaneous masking

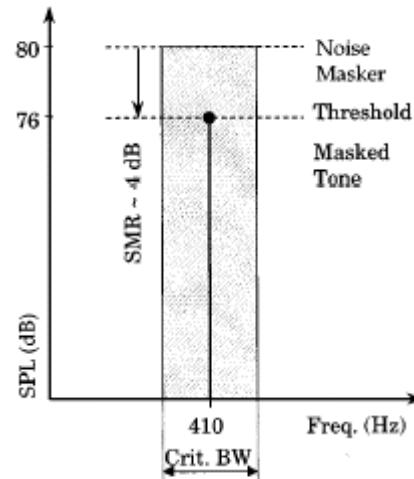
- Pre masking (2 ms)
- Post masking (100 ms)





# Noise Masking Tone

| Filtered Noise<br>Center 410 Hz<br>Width 111 Hz | Tone 1, 820 Hz<br>5 dB below noise | Tone 2, 410 Hz<br>5 dB below noise | Noise<br>+<br>Tone 1 | Noise<br>+<br>Tone 2 |
|---|------------------------------------|------------------------------------|----------------------|----------------------|
|   |                                    |                                    |                      |                      |



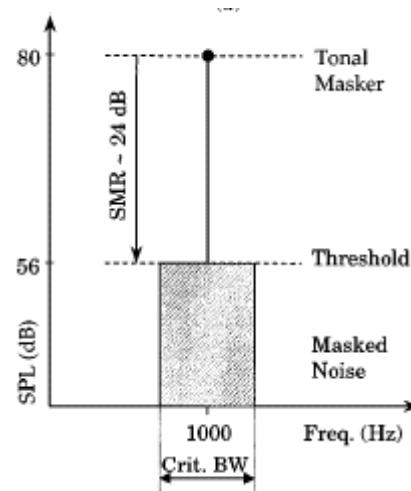
You can not hear a sinusoid that lies in the same critical band as a filtered noise if the sound pressure level is below a certain threshold.

This effect also stretches out beyond the critical band.



# Tone Masking Noise

|   |               |               |                      |                      |
|---|---------------|---------------|----------------------|----------------------|
| Filtered Noise<br>Center 1 kHz<br>Width 162 Hz<br>15 dB below | Tone 1, 2 kHz | Tone 2, 1 kHz | Noise<br>+<br>Tone 1 | Noise<br>+<br>Tone 2 |
|   |               |               |                      |                      |



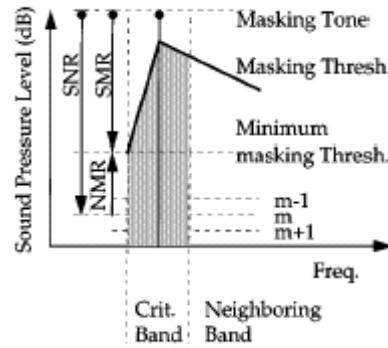
You can not hear a filtered noise that lies in the same critical band as a sinusoid if the sound pressure level is below a certain threshold.

This effect also stretches out beyond the critical band.



# Exploit Masking

- If a sound is masked we can't hear it.



- Make a frequency analysis of the signal and find the masking threshold.
- Put the quantization noise under the masking threshold and we won't hear the quantization.

**Kap 8: DFT, Fast Fourier transform,  
Estimering av effektspektrum**



# Bit Allocation and Masking

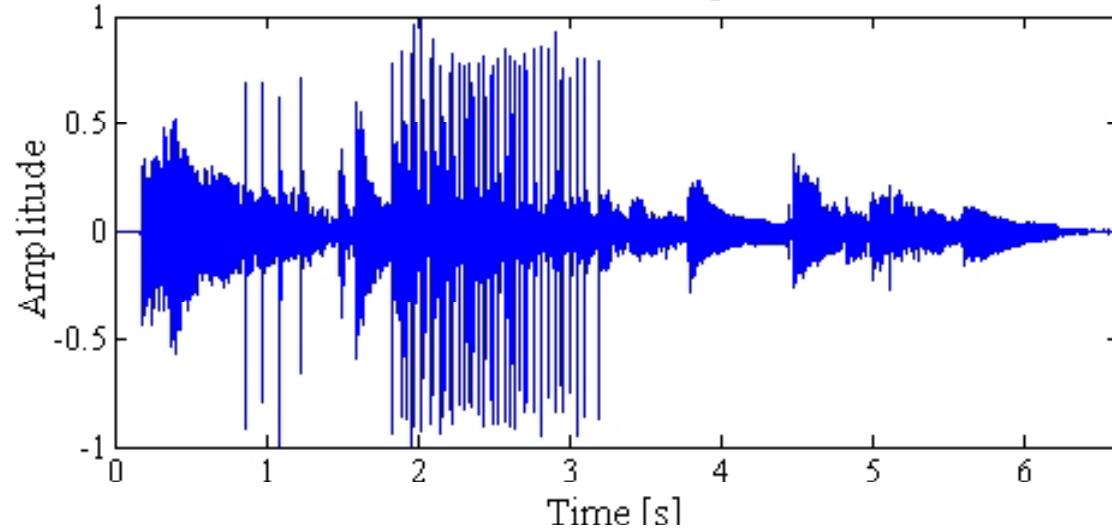
- The masking threshold in each subband gives the Just Noticeable Distortion (JND) limit for that band.
- Bits are assigned to subbands so that the quantization noise falls below or as little over the JND as possible.
- Then the Signal to Quantization noise Ratio (SQR) falls below JND



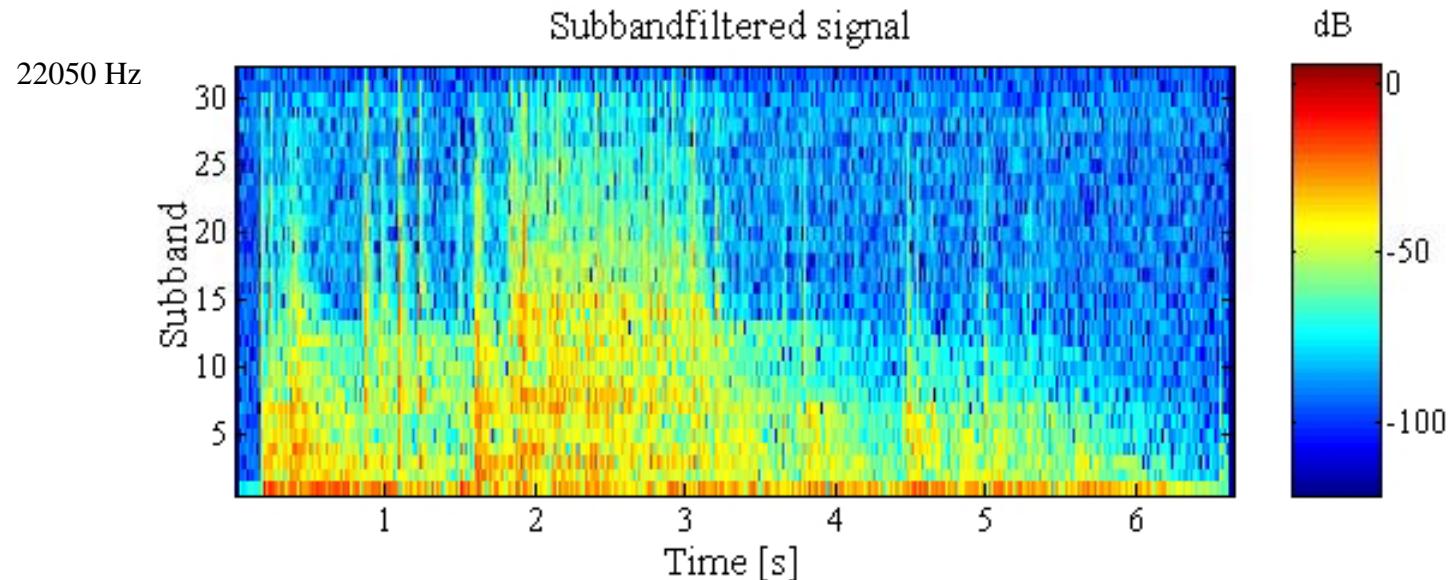
# Castanets and Guitar



Time domain signal

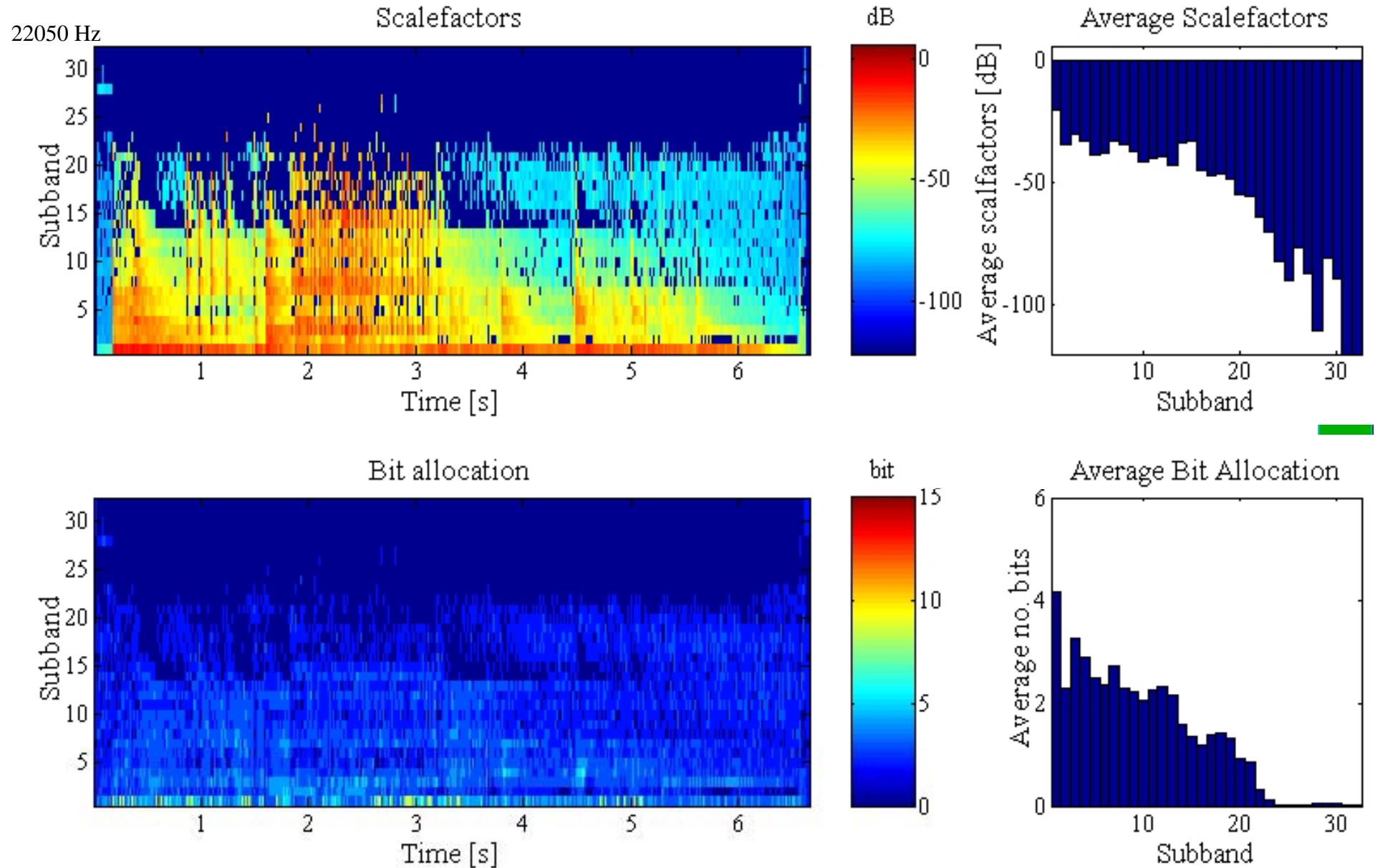


Subbandfiltered signal



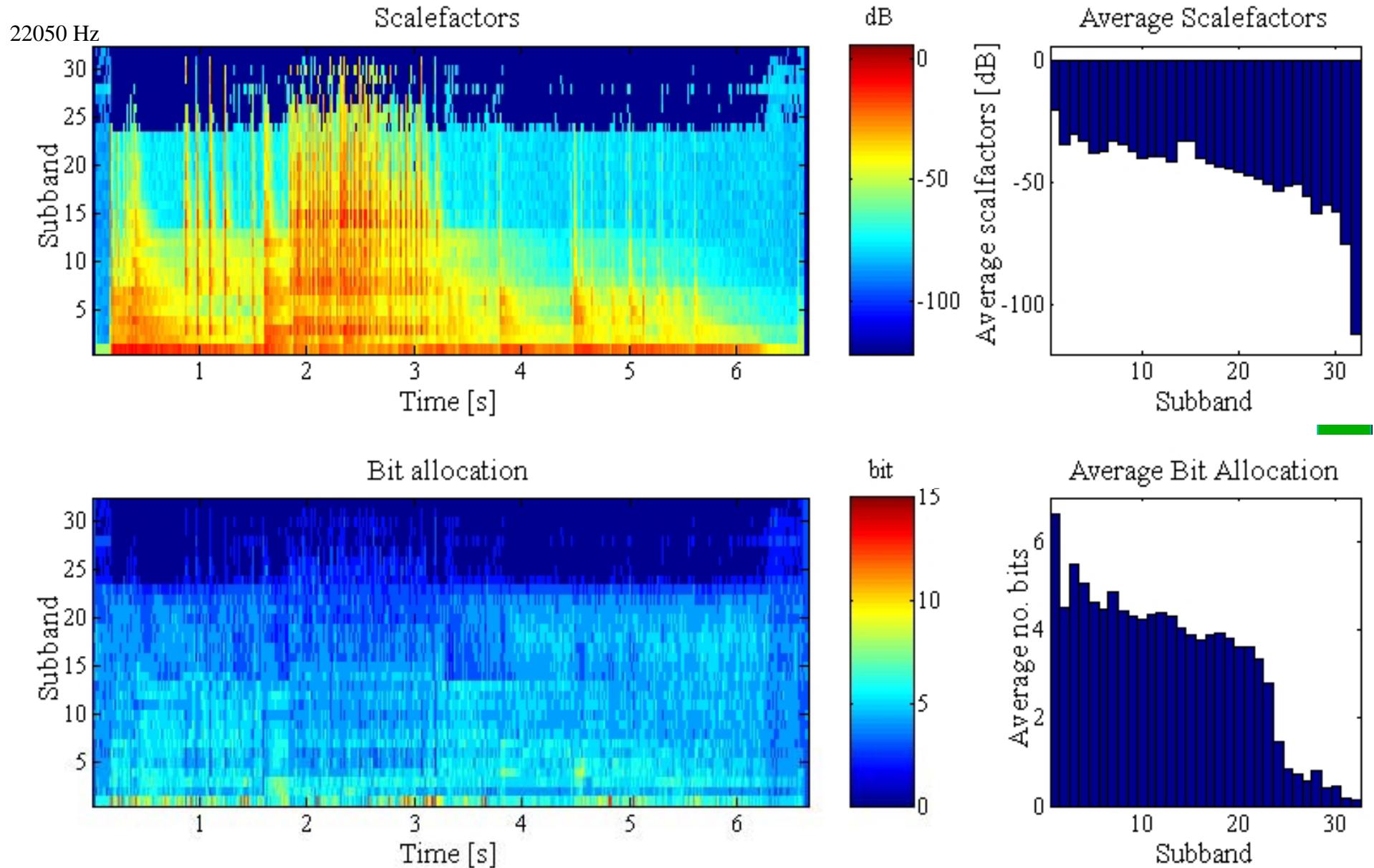


# Bit allocation with 2 bits per sample





# Bit allocation with 4 bits per sample





|    |  |  |
|----|--|--|
| 1  | Overview   | <ul style="list-style-type: none"><li>Praktiske eksempler som mp1/mp3</li></ul>  |
| 2  | Discrete Signals                                     | <ul style="list-style-type: none"><li>Musikk og talesignal</li></ul>   |
| 3  | Time-Domain Analysis                                 | <ul style="list-style-type: none"><li>Linearitet: kan dele i frekvensbånd, behandle for seg og så sette sammen igjen</li><li>Tidsinvarians gjelder bare over kort tid for musikk og tale.</li><li>Differanseligninger: FIR filtre</li></ul>  |
| 4  | z-Transform Analysis                                 | <ul style="list-style-type: none"><li>Analyse av filter i filterbanken: nøkkel til å få til filterdesign</li></ul>   |
| 5  | Frequency Domain Analysis                            | <ul style="list-style-type: none"><li>Frekvensdomene er sentralt i modell av hørsel</li><li>Frekvensselektive filtre: båndpassfiltre</li><li>Inverse systemer: kan dele i bånd i koder og addere sammen igjen i dekoder</li></ul>            |
| 6  | Filter Concepts                                      | <ul style="list-style-type: none"><li>Filterstrukturer, hvordan implementere filterbank i koder og dekoder</li></ul>   |
| 7  | Digital Processing of Analog Signals                 | <ul style="list-style-type: none"><li>A/D-analyse: kvantiseringssstøy ved direkte sampling</li><li>Mutirate system: Hvert delfilter kan nedsamples pga bare 1/32 av total båndbredde =&gt; trenger bare 1/32 samplerate per filter</li></ul> |
| 8  | The Discrete Fourier Transform and Its Applications. | <ul style="list-style-type: none"><li>Frekvensanalyse av signaler</li><li>FFT brukes i estimering av spektrum i koder.</li><li>Må estimere korttidsspektrum for å gjøre adaptiv bittideling</li></ul>  |
| 9  | Design of IIR Filters.                               |  |
| 10 | Design of FIR Filters.                               | <ul style="list-style-type: none"><li>Hvordan finne koeffisienter til bp-filtrene i filterbanken?</li></ul>  |
| 11 | MATLAB Examples                                      |  |
| A  | Useful Concepts from Analog Theory                   |  |