
Adaptive beamforming

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Slide 2: Chapter 7: Adaptive array processing

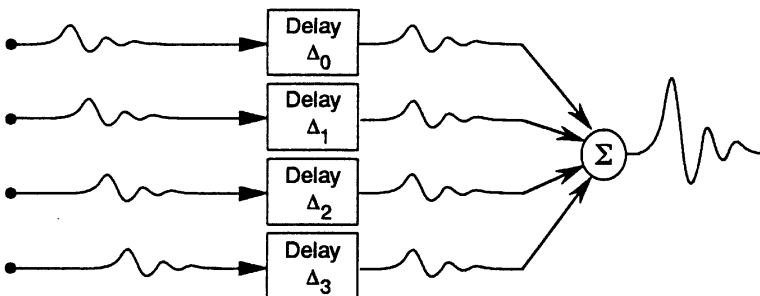
Minimum variance beamforming

- Generalized sidelobe canceler
- Signal coherence
- Spatial smoothing to de-correlate sources

Eigenanalysis methods

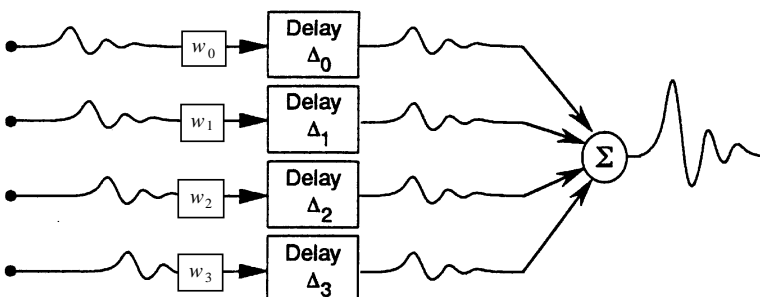
- Signal/noise subspaces
- Eigenvector method
- MUSIC

Slide 3: Delay-and-sum



stacking \triangleq adjustment of $\Delta_0 \dots \Delta_{M-1}$

Slide 4: Delay-and-sum, *continued*



- $z(t) = \sum_{m=0}^{M-1} w_m e^{-j\omega\Delta_m} y_m(t)$
- w_m : weight on signal $m \Rightarrow$ shading = apodization
- Conventional D-A-S: \mathbf{w} independent of recorded signal data

Slide 5: Delay-and-sum on vector form

- Monochromatic source: $y_m(t) = e^{j(\omega t - \vec{k}^\circ \cdot \vec{x}_m)}$.
- Delayed signal: $y_m(t - \Delta_m) = y_m(t) e^{-j\omega\Delta_m}$

Define:

$$\text{Weights: } \mathbf{w} \triangleq \begin{bmatrix} w_0 \\ w_1 \\ \vdots \\ w_{M-1} \end{bmatrix} \quad \text{Received signals: } \widehat{\mathbf{Y}}(t) \triangleq \begin{bmatrix} y_0(t) \\ y_1(t) \\ \vdots \\ y_{M-1}(t) \end{bmatrix} \quad \text{Delayed } \widehat{\mathbf{Y}}(t) : \mathbf{Y}(t) \triangleq \begin{bmatrix} y_0(t)e^{-j\omega\Delta_0} \\ y_1(t)e^{-j\omega\Delta_1} \\ \vdots \\ y_{M-1}(t)e^{-j\omega\Delta_{M-1}} \end{bmatrix}$$

- Beamf. output: $z(t) = \sum_{m=0}^{M-1} w_m e^{-j\omega\Delta_m} y_m(t) = \mathbf{w}^H \mathbf{Y}(t)$

□

- Power of $z(t)$:

$$P(z(t)) \triangleq E\{|z(t)|^2\} = E\{(\mathbf{w}^H \mathbf{Y})(\mathbf{w}^H \mathbf{Y})^H\} = E\{\mathbf{w}^H \mathbf{Y} \mathbf{Y}^H \mathbf{w}\} = \mathbf{w}^H E\{\mathbf{Y} \mathbf{Y}^H\} \mathbf{w} = \mathbf{w}^H \mathbf{R} \mathbf{w}$$

□

Slide 5: Delay-and-sum on vector form, continued

Spatial covariance matrix in case of steering

$$\mathbf{R} \triangleq E\{\mathbf{Y} \mathbf{Y}^H\} = \begin{bmatrix} E\{y_0(t)e^{-j\omega\Delta_0} y_0^*(t)e^{j\omega\Delta_0}\} & E\{y_0(t)e^{-j\omega\Delta_0} y_1^*(t)e^{j\omega\Delta_1}\} & \dots & E\{y_0(t)e^{-j\omega\Delta_0} y_{M-1}^*(t)e^{j\omega\Delta_{M-1}}\} \\ E\{y_1(t)e^{-j\omega\Delta_1} y_0^*(t)e^{j\omega\Delta_0}\} & E\{y_1(t)e^{-j\omega\Delta_1} y_1^*(t)e^{j\omega\Delta_1}\} & \dots & E\{y_1(t)e^{-j\omega\Delta_1} y_{M-1}^*(t)e^{j\omega\Delta_{M-1}}\} \\ \vdots & \vdots & \ddots & \vdots \\ E\{y_{M-1}(t)e^{-j\omega\Delta_{M-1}} y_0^*(t)e^{j\omega\Delta_0}\} & E\{y_{M-1}(t)e^{-j\omega\Delta_{M-1}} y_1^*(t)e^{j\omega\Delta_1}\} & \dots & E\{y_{M-1}(t)e^{-j\omega\Delta_{M-1}} y_{M-1}^*(t)e^{j\omega\Delta_{M-1}}\} \end{bmatrix}$$

Slide 6: Delay-and-sum on vector form, continued

Define:

Steering vector:

$$\mathbf{e} \triangleq \begin{bmatrix} e^{-j\vec{k} \cdot \vec{x}_0} \\ e^{-j\vec{k} \cdot \vec{x}_1} \\ \vdots \\ e^{-j\vec{k} \cdot \vec{x}_{M-1}} \end{bmatrix}$$

- $z(t) = \mathbf{w}^H \mathbf{Y}$
- Power of $z(t)$:

$$P(\mathbf{e}) \triangleq E\{|z(t)|^2\}(\mathbf{e}) = \mathbf{w}^H \mathbf{R}(\mathbf{e}) \mathbf{w} \stackrel{\text{if } w_m=1, \forall m}{=} \mathbf{1}^H \mathbf{R}(\mathbf{e}) \mathbf{1}$$

Slide 7: Delay-and-sum on vector form, *continued*

About \mathbf{e} :

- Steering vector

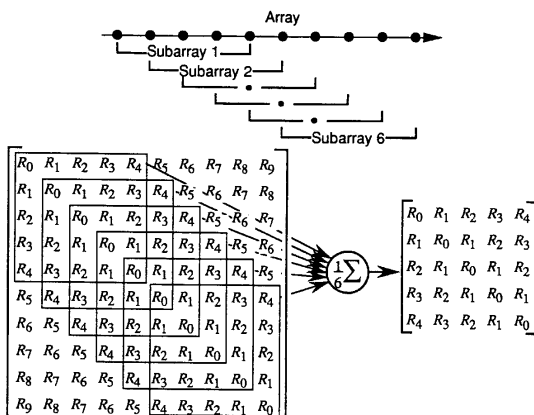
$$\mathbf{e} \triangleq \begin{bmatrix} e^{-j\vec{k}\cdot\vec{x}_0} \\ e^{-j\vec{k}\cdot\vec{x}_1} \\ \vdots \\ e^{-j\vec{k}\cdot\vec{x}_{M-1}} \end{bmatrix}$$

- Contains delays to focus in specific direction
- Represents unit amplitude signal, propagating in \vec{k} direction

In conventional D-A-S: \mathbf{w} independent of received signal data

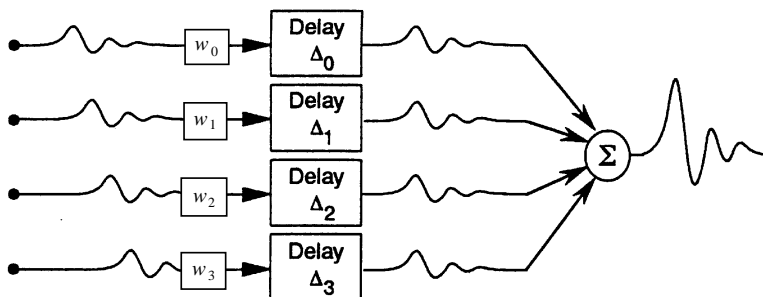
Slide 8: Estimation of spatial covariance matrix

- Averaging in time
- Averaging in space = spatial smoothing = subarray averaging



Slide 9: Minimum variance beamforming

Assume narrow-band signals



- Adaptive method. [Latin: *adaptare* “to fit to”]
- Allow w_m to also be complex and/or negative
- Sensor weights defined not only as function of problem geometry, but also as function of received signals

Slide 10: Minimum variance beamforming, *continued*

- “Minimum variance” = “Capon” = “Maximum likelihood” (beamforming)
- Constrained optimization problem
- New steering direction $\mathbf{e} \Rightarrow$ new calculation of element weights w_m

□

Minimum-variance key equations

$$\mathbf{w} \triangleq \begin{bmatrix} w_0 \\ \vdots \\ w_{M-1} \end{bmatrix} \Rightarrow \begin{array}{l} \min_{\mathbf{w}} (\mathbf{w}^H \mathbf{R} \mathbf{w}), \\ \text{constraint:} \\ \mathbf{w}^H \mathbf{e} = 1 \end{array} \Leftrightarrow \mathbf{w} = \frac{\mathbf{R}^{-1} \mathbf{e}}{\mathbf{e}^H \mathbf{R}^{-1} \mathbf{e}} \Rightarrow P(\mathbf{e}) = \frac{1}{\mathbf{e}^H \mathbf{R}^{-1} \mathbf{e}}$$

- [beampattern] \neq [steered response]

Slide 11: Minimum variance beamforming, *continued*

Example

Uniform linear array, $M = 10$

Comparing:

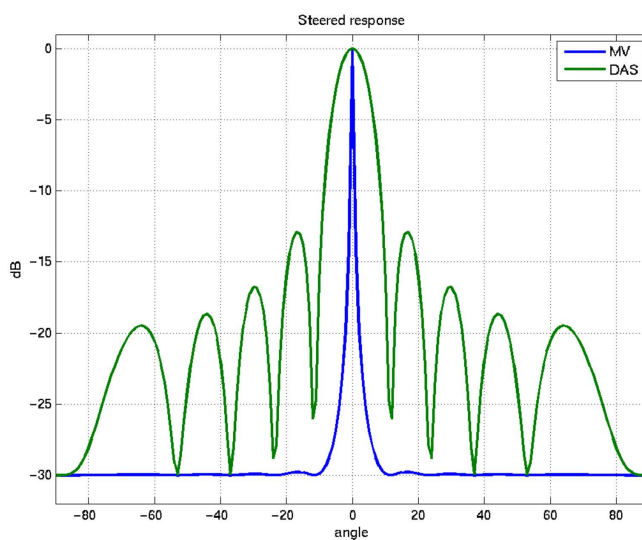
- Delay-and-sum
- Minimum-variance

Investigating:

- Steered responses
- Beampatterns

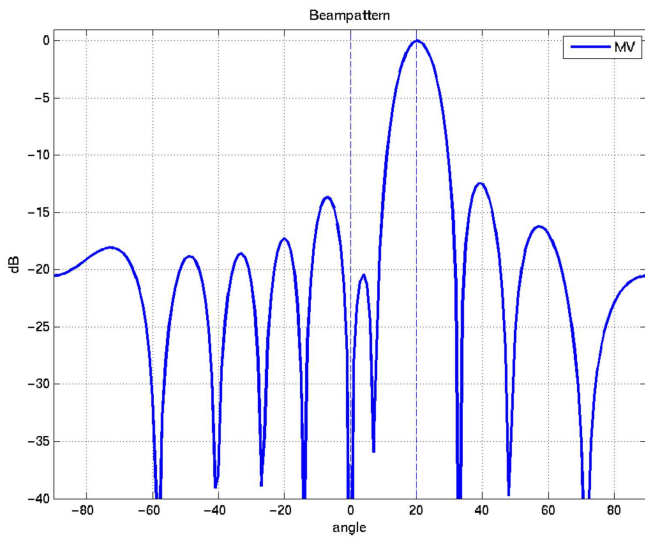
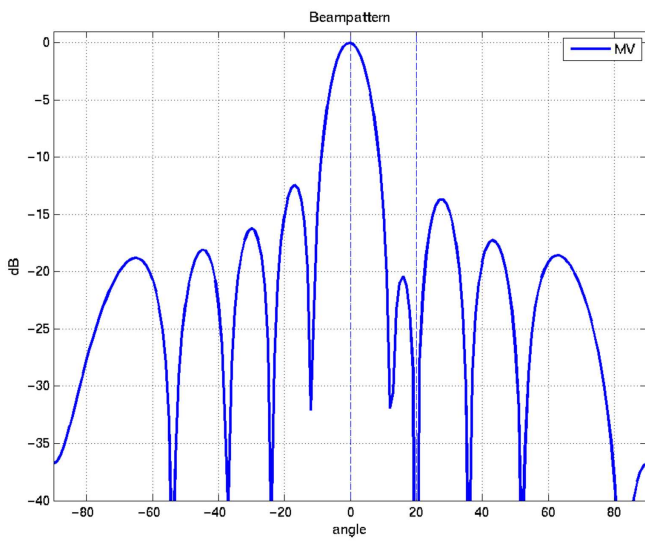
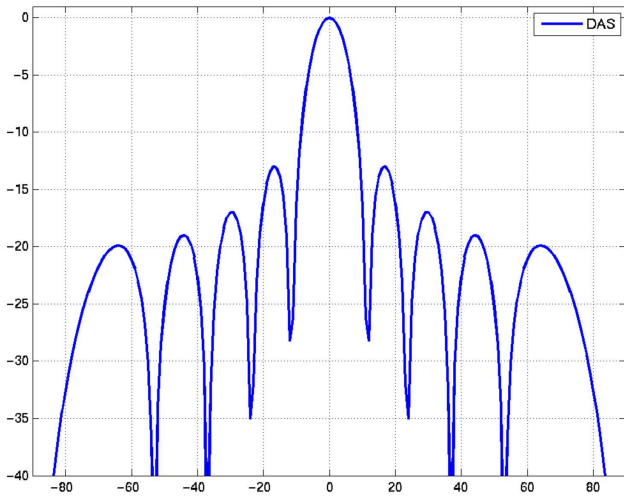
Slide 12: Comparison: delay-and-sum / minimum-variance

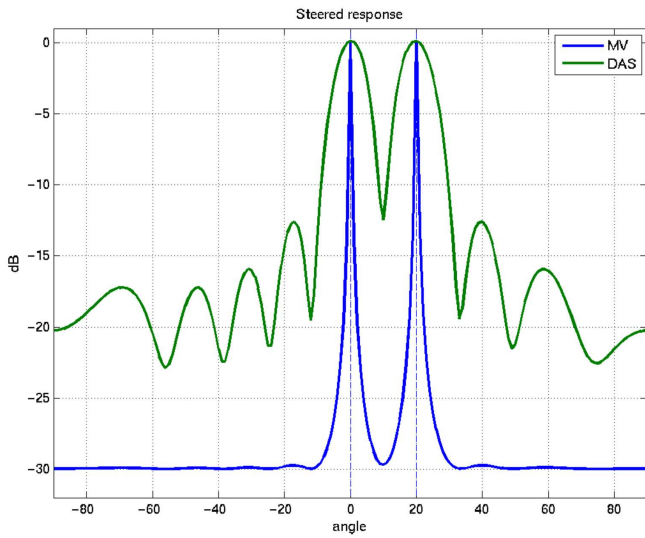
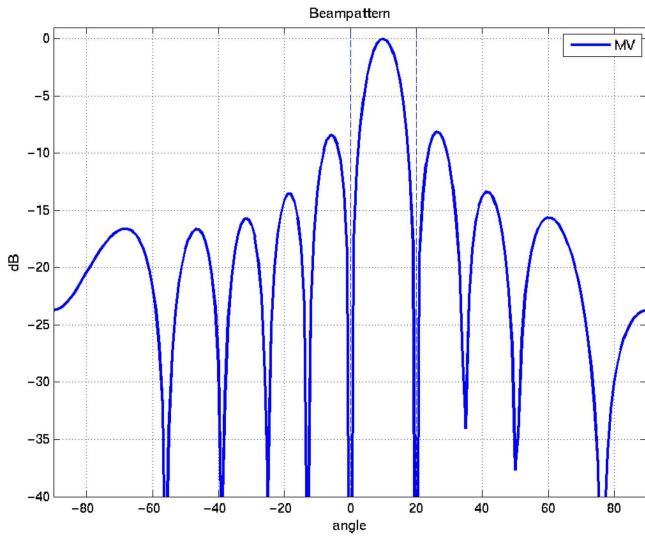
1 source



Slide 13: Comparison: delay-and-sum / minimum-variance

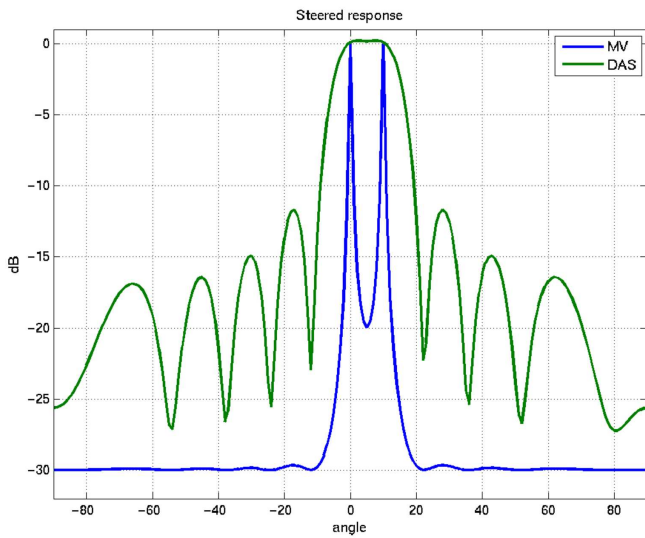
2 sources





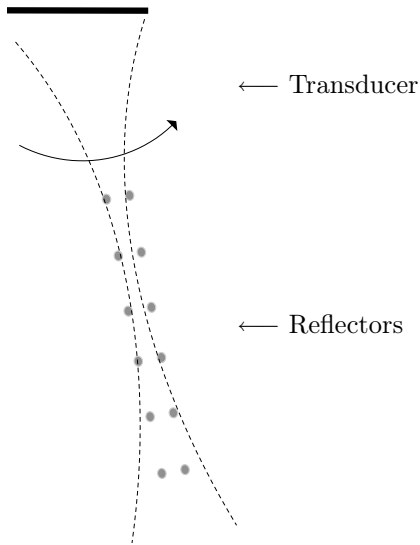
Slide 14: Comparison: delay-and-sum / minimum-variance

2 closely spaced sources



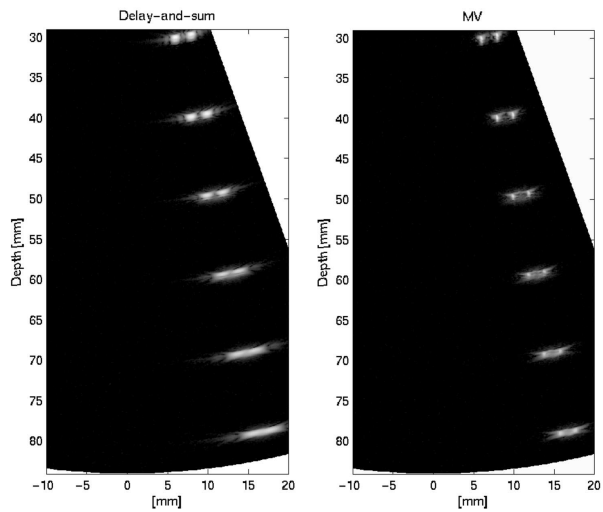
Slide 15: Example: Ultrasound imaging

Experimental setup



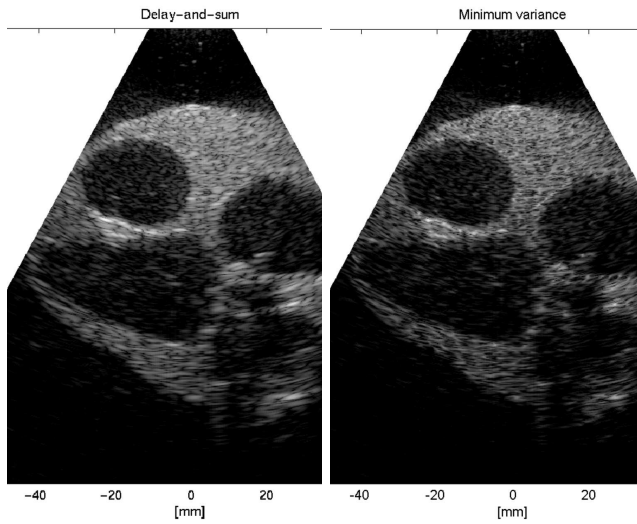
Slide 16: Example: Ultrasound imaging, *continued*

Resulting images

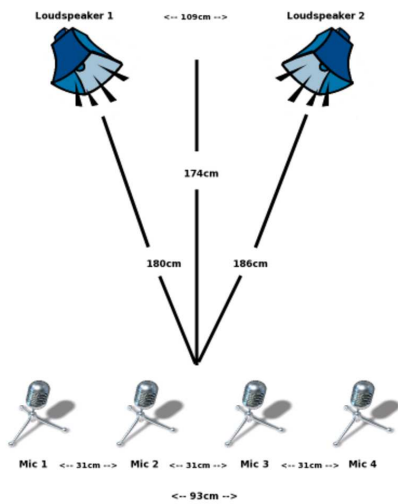


Slide 17: Example: Ultrasound imaging, *continued*

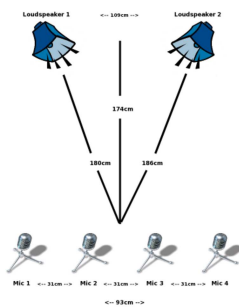
Experimental data, heart phantom



Slide 18: Sound example: Microphone array



Slide 19: Sound example, *continued*



- Single microphone 🗣️ mix.wav
- Delay-and-sum 🗣️ gaute_DAS.wav

- Minimum-variance  gaute_adaptive.wav

Slide 20: Minimum-variance beamformer:

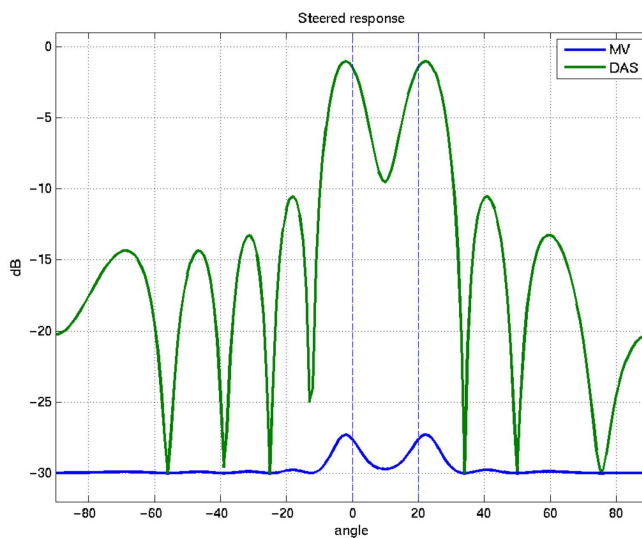
Uncorrelated signals required for optimum functioning



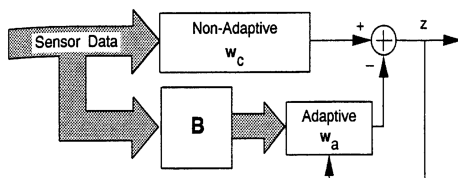
- Correlated signals may give signal cancellation
- Although constraint is fulfilled: signal propagating in the “look-direction” may be canceled by a correlated interferer

Slide 21: Minimum-variance / D-A-S example

2 correlated sources



Slide 22: Generalized sidelobe canceler



- Conventional D-A-S weights w_c : steering in assumed propagation direction
- Adaptive/canceler portion: removal of other signals
- Blocking matrix \mathbf{B} , blocks signal from assumed propagation direction from coming into “canceler part”
- w_a : adaptive weights to emphasize what is to be removed
- Minimize total power. Unconstrained minimization:

$$\min_{w_a} (w_c - \mathbf{B}^H w_a)^H \mathbf{R} (w_c - \mathbf{B}^H w_a) \Rightarrow w_a = (\mathbf{B}\mathbf{R}\mathbf{B}^H)^{-1} \mathbf{B}\mathbf{R}w_c$$

- Full details: D&J pp. 369–371

Slide 23: Eigenanalysis and Fourier analysis

(Chapter 7.3.1)

Eigenvalues / Eigenvectors of \mathbf{R}

$$\mathbf{R}\mathbf{v}_i = \lambda_i\mathbf{v}_i$$

$$\left. \begin{array}{l} \mathbf{R} \text{ Hermitian (self-adjoint)} \Leftrightarrow \mathbf{R} = \mathbf{R}^H \\ \mathbf{R} \text{ Positive semidefinite} \Leftrightarrow \mathbf{x}^H \mathbf{R} \mathbf{x} \geq 0, \forall \mathbf{x} \neq \mathbf{0} \end{array} \right\} \Rightarrow$$

- Eigenvalues real & positive: $\lambda_i \geq 0, \forall i$
- Eigenvectors \mathbf{v}_i : orthogonal

$$\Rightarrow \text{For } \mathbf{V} \triangleq \begin{bmatrix} | & | & \dots & | \\ \mathbf{v}_1 & \mathbf{v}_2 & \dots & \mathbf{v}_M \\ | & | & \dots & | \end{bmatrix}: \mathbf{V} \text{ is unitary matrix}$$

$$\Leftrightarrow \mathbf{V}^H \mathbf{V} = \mathbf{I} \Leftrightarrow \boxed{\mathbf{V}^{-1} = \mathbf{V}^H}$$

□

Slide 24: Eigenvalue decomposition

□

Eigendecomposition of \mathbf{R} & \mathbf{R}^{-1} : *The spectral theorem*

- $\mathbf{R} = \sum_{i=1}^M \lambda_i \mathbf{v}_i \mathbf{v}_i^H = \mathbf{V} \mathbf{\Lambda} \mathbf{V}^H$
- $\mathbf{R}^{-1} = \sum_{i=1}^M \frac{1}{\lambda_i} \mathbf{v}_i \mathbf{v}_i^H = \mathbf{V} \mathbf{\Lambda}^{-1} \mathbf{V}^H$
- *Note above:* Same eigenvectors, but inverse eigenvalues

Slide 25: [Signal + noise] & [noise] subspaces

$$\mathbf{R} = \mathbf{V} \mathbf{\Lambda} \mathbf{V}^H = \underbrace{\mathbf{V}_s \mathbf{\Lambda}_s \mathbf{V}_s^H}_{\substack{\text{from} \\ \text{signal}}} + \underbrace{\mathbf{V}_n \mathbf{\Lambda}_n \mathbf{V}_n^H}_{\substack{\text{from} \\ \text{noise}}}$$

□

Slide 26: Beamformer output powers

□

Minimum variance, $\hat{\mathbf{R}} = \mathbb{E}\{\mathbf{Y}\mathbf{Y}^H\}$: “full \mathbf{R} estimate”

$$P_{\text{MV}}(\mathbf{e}) = \frac{1}{\underbrace{\mathbf{e}^H \mathbf{V}_s \mathbf{\Lambda}_s^{-1} \mathbf{V}_s^H \mathbf{e}}_{\text{signal-plus-noise subspace}} + \underbrace{\mathbf{e}^H \mathbf{V}_n \mathbf{\Lambda}_n^{-1} \mathbf{V}_n^H \mathbf{e}}_{\text{noise subspace}}} = \frac{1}{\sum_{i=1}^{N_s} \frac{1}{\lambda_i} |\mathbf{e}^H \mathbf{v}_i|^2 + \sum_{i=N_s+1}^M \frac{1}{\lambda_i} |\mathbf{e}^H \mathbf{v}_i|^2}$$

Eigenvector method, $\hat{\mathbf{R}} = \sum_{i=N_s+1}^M \lambda_i \mathbf{v}_i \mathbf{v}_i^H$: “noise-only \mathbf{R} estimate”

$$P_{\text{EV}}(\mathbf{e}) = \frac{1}{\mathbf{e}^H \mathbf{V}_n \mathbf{\Lambda}_n^{-1} \mathbf{V}_n^H \mathbf{e}} = \frac{1}{\sum_{i=N_s+1}^M \frac{1}{\lambda_i} |\mathbf{e}^H \mathbf{v}_i|^2}$$

MUSIC, $\hat{\mathbf{R}} = \sum_{i=N_s+1}^M \mathbf{v}_i \mathbf{v}_i^H$: “normalized noise-only \mathbf{R} estimate”

$$P_{\text{MUSIC}}(\mathbf{e}) = \frac{1}{\mathbf{e}^H \mathbf{V}_n \mathbf{V}_n^H \mathbf{e}} = \frac{1}{\sum_{i=N_s+1}^M |\mathbf{e}^H \mathbf{v}_i|^2}$$

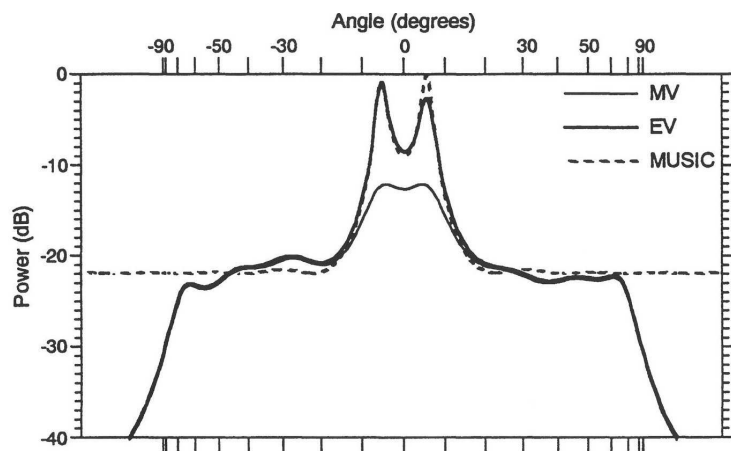
Slide 27: [Signal + noise], & [noise] subspaces, *continued*

- \mathbf{R} always Hermitian $\Rightarrow \forall$ eigenvectors orthogonal
- N_s largest eigenvectors: span the [signal+noise subspace]
- $M - N_s$ smallest span the [noise subspace]
- Signal steering vectors \in [signal+noise subspace], \perp [noise subspace]
- [Largest eigenvectors] \neq [signal vectors] However: signal vectors are linear combinations of [largest eigenvectors]

Slide 28: EV / MUSIC

- (Inverse of) projection of possible steering vectors onto the noise subspace
- MUSIC: Dropping λ_i in $\hat{\mathbf{R}} \Leftrightarrow$ noise whitening
- EV & MUSIC: peaks at Directions-Of-Arrival (DOAs)
 - No amplitude preservation
 - Require knowledge of # signals, N_s
 - How to estimate N_s ?

Slide 29: MV / EV / MUSIC comparison



Slide 30: Situation with signal coherence

□

$$\mathbf{R} = \mathbf{S}\mathbf{C}\mathbf{S}^H + \mathbf{K}_n$$

- signal coherence \Rightarrow rank deficient signal matrix $\mathbf{S}\mathbf{C}\mathbf{S}^H$
- For minimum-variance: May cause signal cancellation
- Gives signal+noise-subspace consisting of linear combinations of signal vectors
- EV & MUSIC: may fail to produce peaks at DOA locations
- Want to reduce cross-correlation terms in \mathbf{C} .
- Cure: Spatial smoothing (subarray averaging)