

# Transistor brukt som forsterker

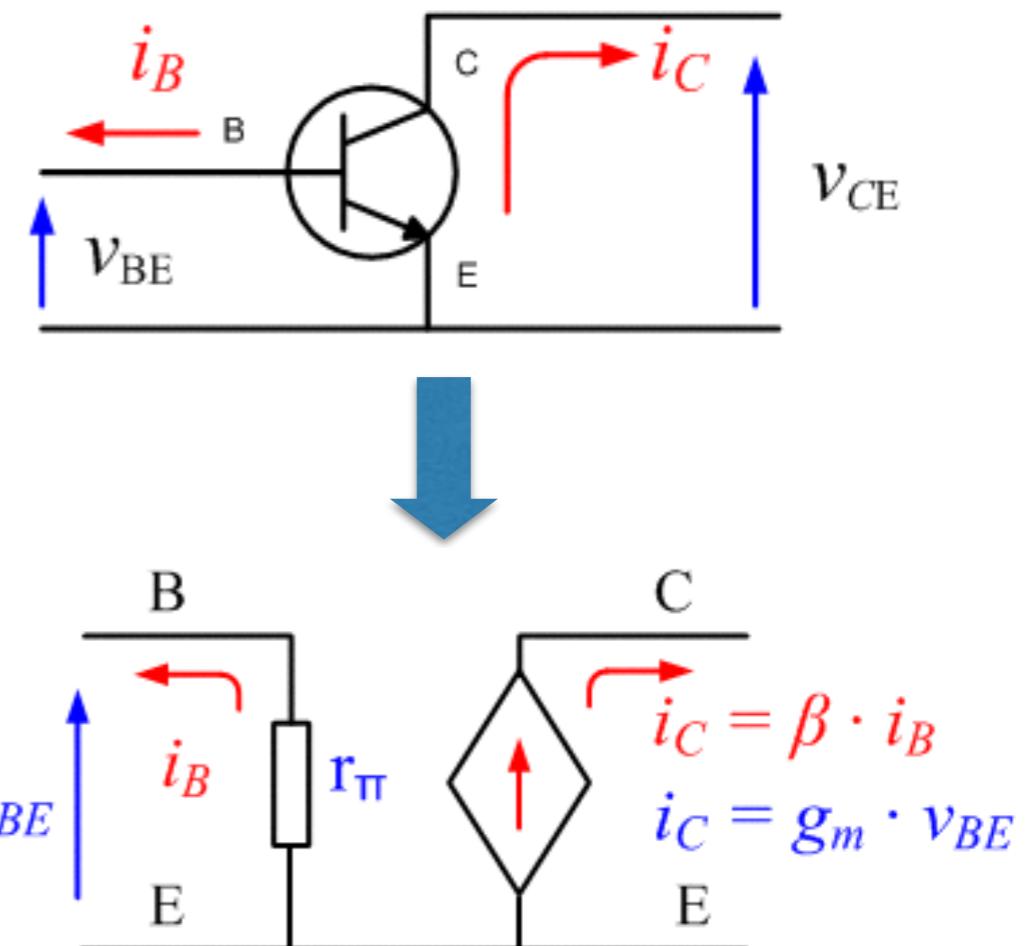
## Vi ser på Småsignalmodeller

Vi har sett hvordan vi vha. en emittermotstand kan stabilisere forsterkerens arbeidspunkt

- Alle betrakninger så langt er gjort med en DC – modell av forsterkeren. ( En statisk beregningsmodell )

Men hvordan virker forsterkeren for små signaler?

Vi erstatter det vanlige transistorsymbolet med en småsignalmodell og signalstrømmer og spenninger angis med små bokstaver



Mellan Base og Emitter "ser" signalet en "dynamisk" motstand  $r_\pi$  (BE-dioden).

Mellan Emitter og Collector finner vi en strømgenerator som leverer signalstrømmen  $i_C$ . Denne strømmen bestemmes av transistorens transkonduktans  $g_m$   
 $r_\pi$  og  $g_m$  kalles småsignalparametere

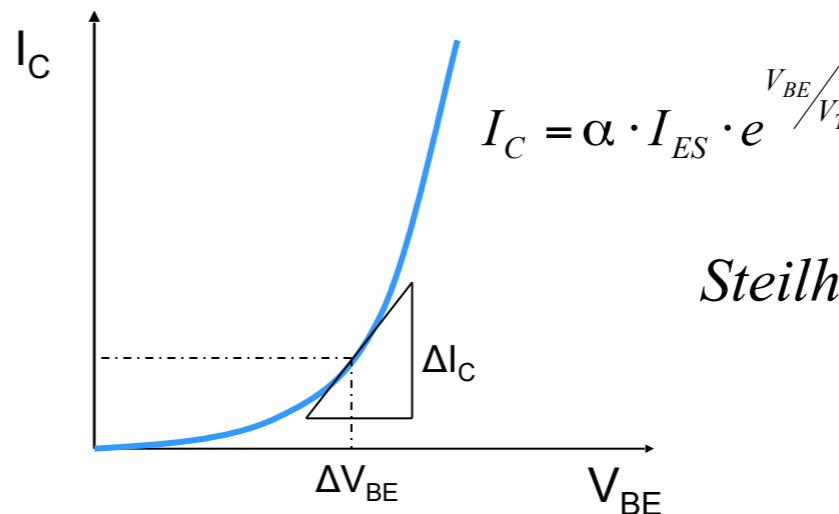
# Småsignalparameterer

Transkonduktans - steilhet  $g_m$

$$\text{Emitterstrømmen } I_E = I_{ES} \cdot e^{\frac{V_D}{V_T}}$$

hvor  $V_D = V_{BE}$  og  $V_T = 25mV$   
(diodelikningen)

$$I_C = I_E - I_B \rightarrow I_C = \alpha \cdot I_E \quad \alpha \approx 1$$



$$\text{Steilhet } g_m = \frac{\Delta I_C}{\Delta V_{BE}}$$

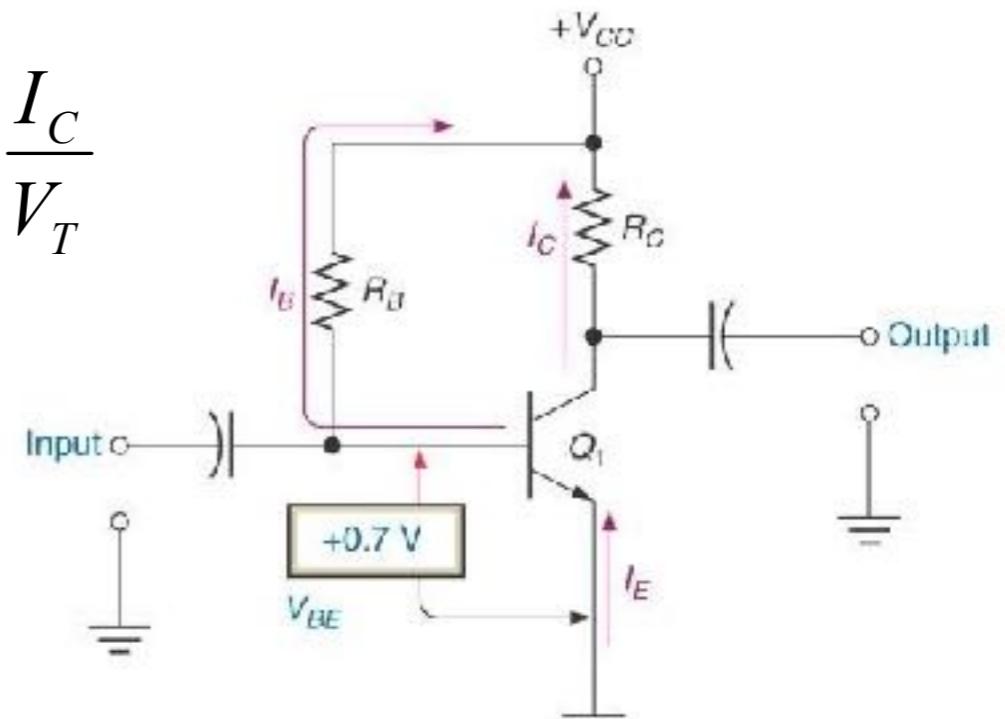
Steilheten  $g_m$  er gitt av tangenten til kurven for  $I_C$ . Deriverer  $I_C$  mhp.  $V_{EB}$

$$g_m = \frac{d(I_C)}{dV_{EB}} = \alpha \cdot I_{ES} \cdot e^{\frac{V_{EB}}{V_T}} \cdot \frac{1}{V_T} = I_C \cdot \frac{1}{V_T} = \frac{I_C}{V_T}$$

$$g_m = \frac{I_C}{V_T}$$

Eksempel : Forsterkeren settes opp med  $I_C = 2 \text{ mA}$ , som gir:

$$g_m = \frac{I_C}{V_T} = \frac{2mA}{25mV} = 80mS \quad (\text{benevning Siemens})$$



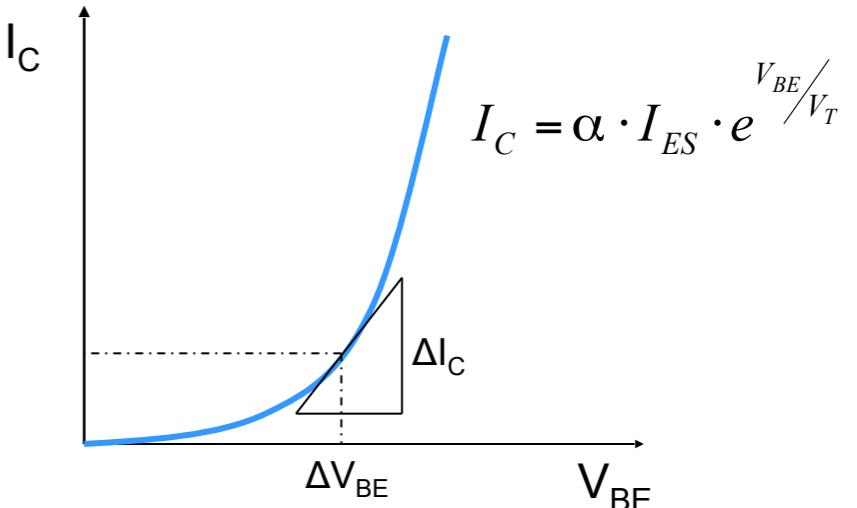
# Småsignalparameter

Dynamisk inngangsmotstand  $r_\pi$

$$r_\pi = \frac{\Delta V_{EB}}{\Delta I_B}$$

$$I_C = \beta \cdot I_B$$

en liten endring i  $I_B$  gir stor endring i  $I_C$

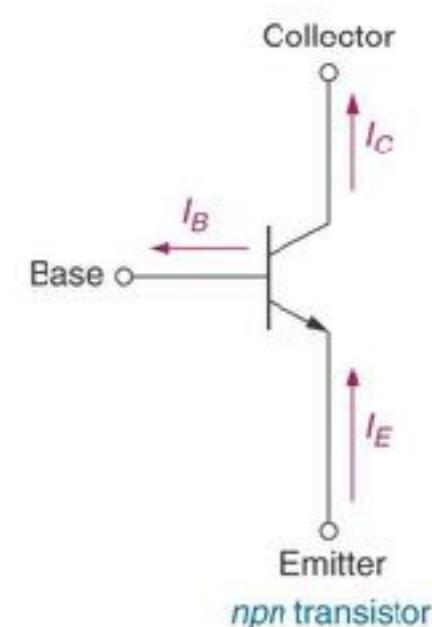


$$1) \Delta I_B = \frac{\Delta I_C}{\beta} \quad 2) g_m = \frac{\Delta I_C}{\Delta V_{EB}} \rightarrow \Delta I_C = g_m \cdot \Delta V_{EB}$$

Forholdet mellom  $\Delta V_{EB}$  og  $\Delta I_B$  kalles  
den dynamiske inngangsresistansen  $r_\pi$   
Kombinerer likning 1) og 2)

$$\Delta I_B = \frac{g_m \cdot \Delta V_{EB}}{\beta}$$

$$r_\pi = \frac{\Delta V_{EB}}{\Delta I_B} = \frac{\beta}{g_m} = \frac{\beta \cdot V_T}{I_C}$$



# Transistorforsterker

Vi beregner spenningsforsterkningen  $A_V$

$$Steilhet \ g_m = \frac{\Delta I_C}{\Delta V_{BE}}$$

$$1) \ \Delta I_C = g_m \cdot \Delta V_{BE}$$

$$2) \ \Delta V_{RC} = \Delta I_C \cdot R_C \quad (ohms lov)$$

Setter inn 1) i 2) som gir

$$\Delta V_{RC} = g_m \cdot \Delta V_{BE} \cdot R_C$$

Forsterningen  $A_V$  er definert som

$$A_V = \frac{V_{Output}}{V_{Input}} = \frac{\Delta V_{RC}}{\Delta V_{BE}} = g_m \cdot R_C$$

Gitt  $V_{CC}=10\text{volt}$  Setter  $V_C=5\text{volt}$  Vi bestemmer at  $I_C = 2\text{mA}$

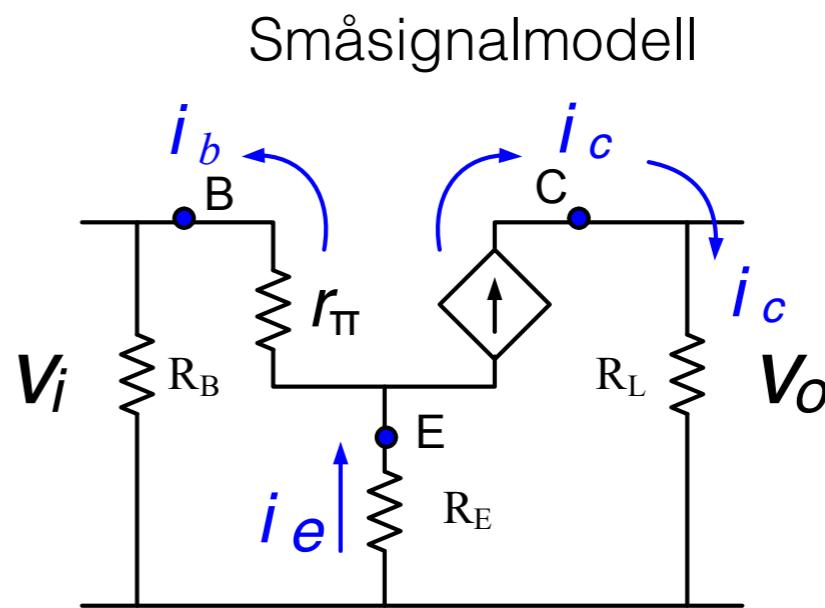
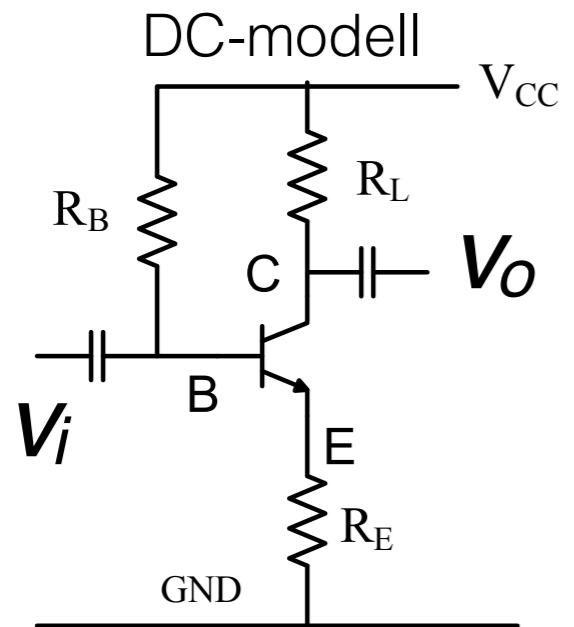
$$A_V = g_m \cdot R_C$$

Beregner  $R_C = \frac{V_{RC}}{I_C} = \frac{5v}{2mA} = 2,5 k\Omega$      $g_m = \frac{I_C}{V_T} = \frac{2mA}{25mV} = 80mS$

Forsterkningen  $A_V = 80 mS \cdot 2,5 k\Omega = 200$

# Transistorforsterker

Småsignalmodeller- forsterker med emittermotstand



$$v_i = v_{be} + v_E \quad (1) \qquad v_{be} = i_b \cdot r_\pi \quad (2)$$

$$v_E = (i_b + i_c) \cdot R_E = (i_b + i_b \cdot \beta) \cdot R_E = i_b (\beta - 1) \cdot R_E \quad (3)$$

Setter (2) og (3) inn i (1)

$$v_i = i_b \cdot r_\pi + i_b (\beta + 1) \cdot R_E \rightarrow i_b = \frac{v_i}{r_\pi + (\beta + 1) \cdot R_E} \quad (4)$$

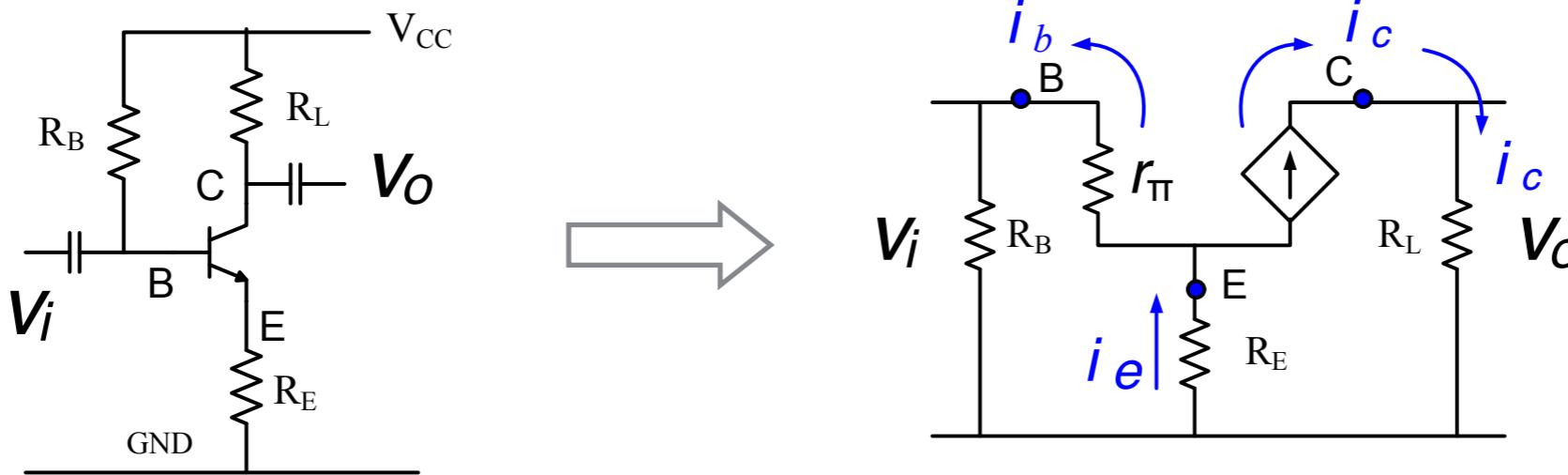
$$v_o = -i_c \cdot R_L = -i_b \cdot \beta \cdot R_L \rightarrow (4) \rightarrow = -\frac{\beta \cdot v_i}{r_\pi + (\beta + 1) \cdot R_E} \cdot R_L \qquad A_v = -\frac{v_o}{v_i}$$

Bemerk (-) Forsterkeren inverterer signalet.

Positivt signal-sving på basen kommer ut som negativt sving på kollektor

# Transistorforsterker

Småsignalmodeller- forsterker med emittermotstand



$$v_i = i_b \cdot r_\pi + i_b(\beta + 1) \cdot R_E \rightarrow i_b = \frac{v_i}{r_\pi + (\beta + 1) \cdot R_E} \quad (4)$$

$$v_o = -i_c \cdot R_L = -i_b \cdot \beta \cdot R_L \rightarrow (4) \Rightarrow = -\frac{\beta \cdot v_i}{r_\pi + (\beta + 1) \cdot R_E} \cdot R_L$$

$$A_v = \frac{v_o}{v_i} = -\frac{\beta \cdot R_L}{r_\pi + (\beta + 1) \cdot R_E} \quad \text{setter inn } r_\pi = \frac{\beta}{g_m}$$

$$A_v = -\frac{g_m \cdot R_L}{1 + (1 + \frac{1}{\beta}) \cdot g_m \cdot R_E} \approx -\frac{g_m \cdot R_L}{1 + g_m \cdot R_E} \approx -\frac{R_L}{R_E} \quad \Rightarrow$$

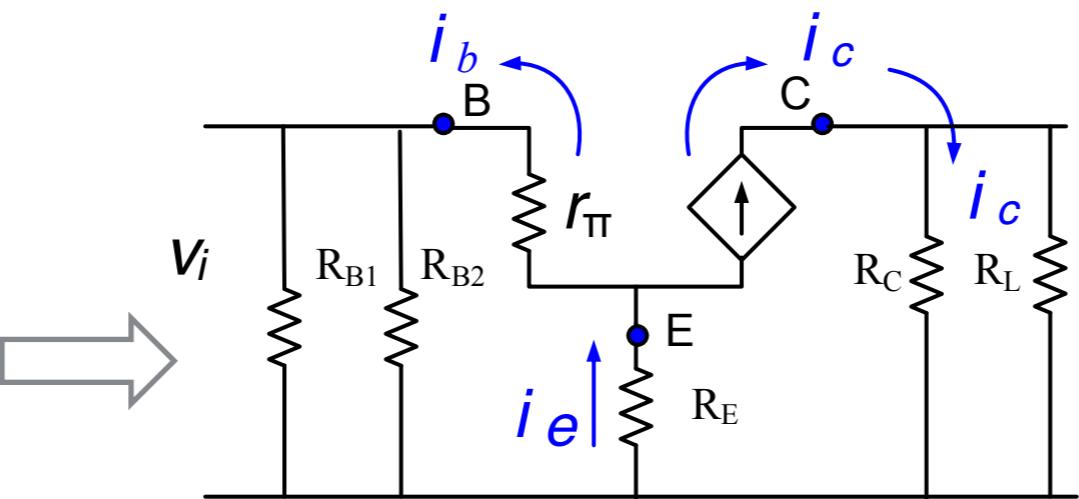
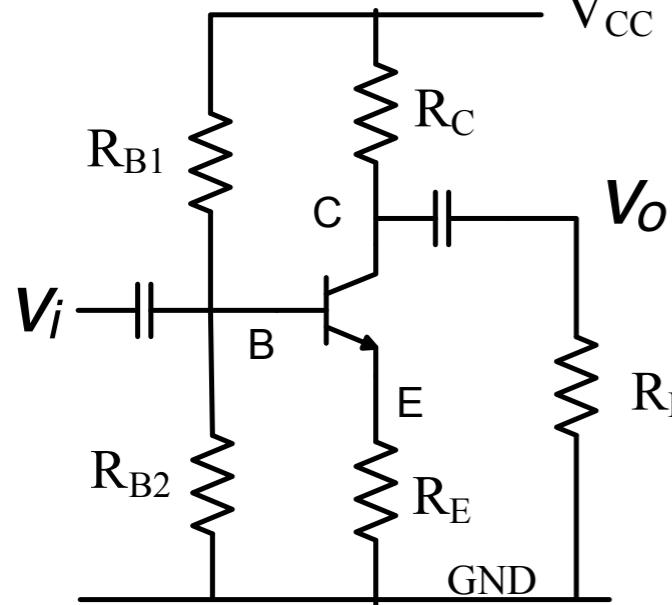
$$A_v \approx -\frac{R_L}{R_E}$$

Ser på "input resistance" til transistoren  $r_{inn} = \frac{v_i}{i_b} \rightarrow fra (4) \rightarrow \underline{\underline{r_\pi + (\beta + 1)R_E}}$

# Transistorforsterker

Småsignalmodeller- forsterker med emittermotstand

Forsterker med emittermotstand



Forsterkning

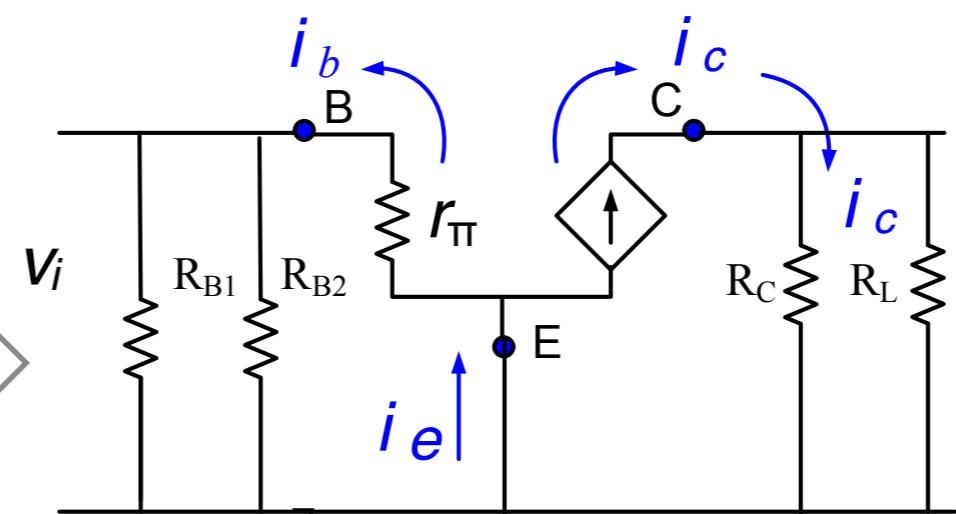
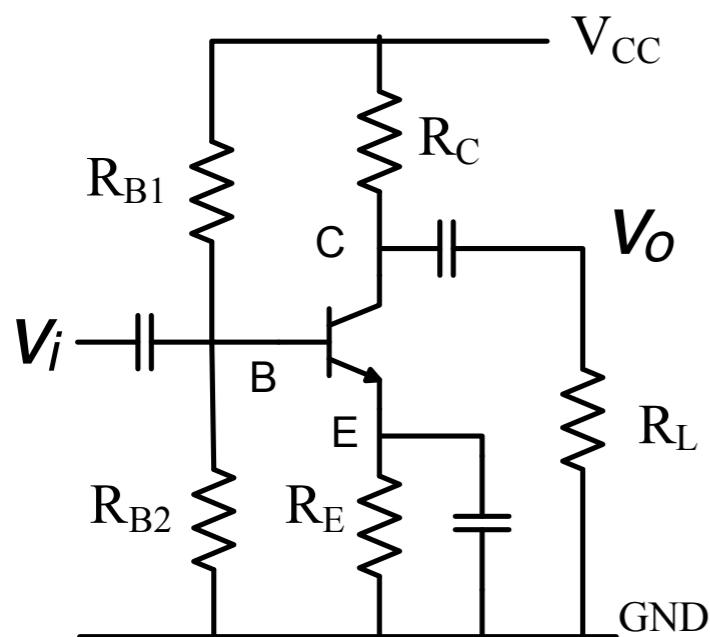
$$A_v \approx -\frac{R_C \parallel R_L}{R_E}$$

Signalkilden ser inn mot en motstand

$$R = R_{B1} \parallel R_{B2} \parallel r_{inn} \text{ hvor}$$

$$r_{inn} = r_\pi + (\beta + 1)R_E$$

Forsterker med avkoplet emittermotstand



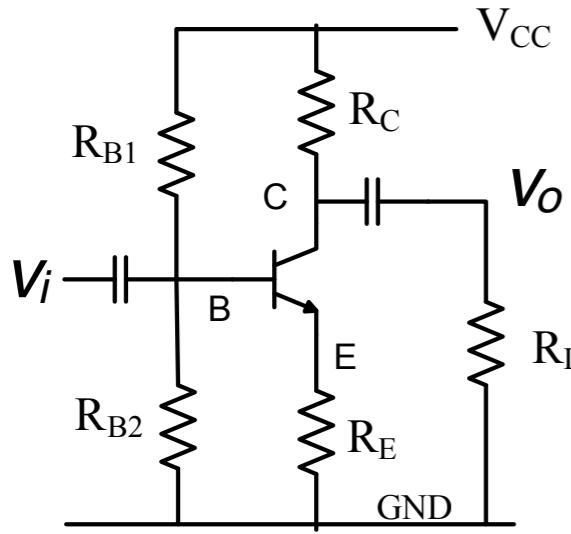
Forsterkning

$$A_V = -g_m \cdot (R_C \parallel R_L)$$

$$g_m = \frac{I_C}{V_T}$$

# Transistorforsterker

## Regneeksempel



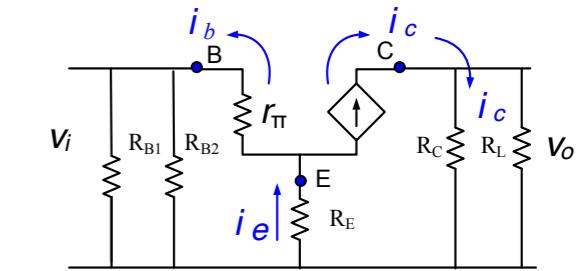
$$R_{B1} = 35 \text{ k}\Omega, R_{B2} = 15 \text{ k}\Omega, R_C = 6 \text{ k}\Omega,$$

$$R_L = 1,5 \text{ k}\Omega, R_E = 1 \text{ k}\Omega, I_C = 2 \text{ mA}, \beta = 200$$

$$g_m = \frac{I_C}{V_T} = \frac{2 \text{ mA}}{25 \text{ mV}} = 80 \text{ mS} \quad r_\pi = \frac{\beta}{g_m} = \frac{200}{80 \text{ mS}} = 2,5 \text{ k}\Omega$$

$$R_P = R_C \parallel R_L = \frac{6 \text{ k} \cdot 1,5 \text{ k}}{6 \text{ k} + 1,5 \text{ k}} = 1,2 \text{ k} \quad A_V = -\frac{R_P}{R_E} = \frac{1,2 \text{ k}}{1 \text{ k}} = -1,2$$


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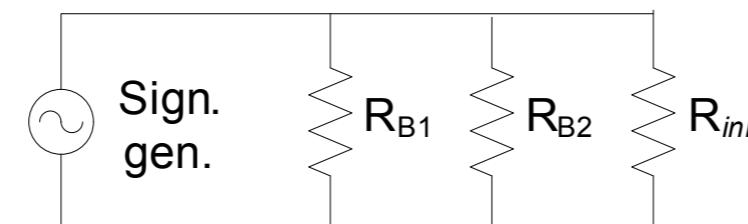
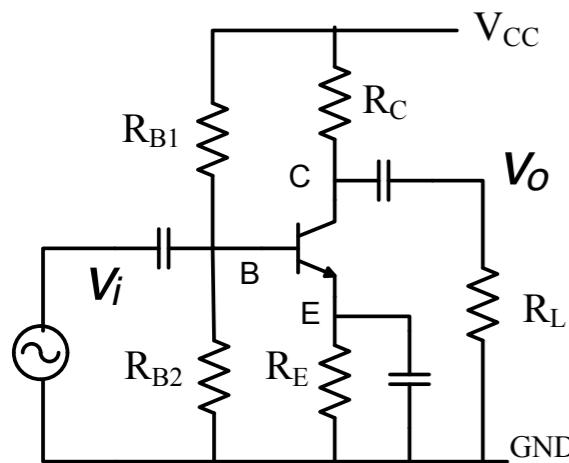


Hvis vi avkopler emittermotstanden med en kondensator

$$\underline{A_v = -g_m \cdot R_P = 80 \text{ mS} \cdot 1,2 \text{ k} = -96}$$

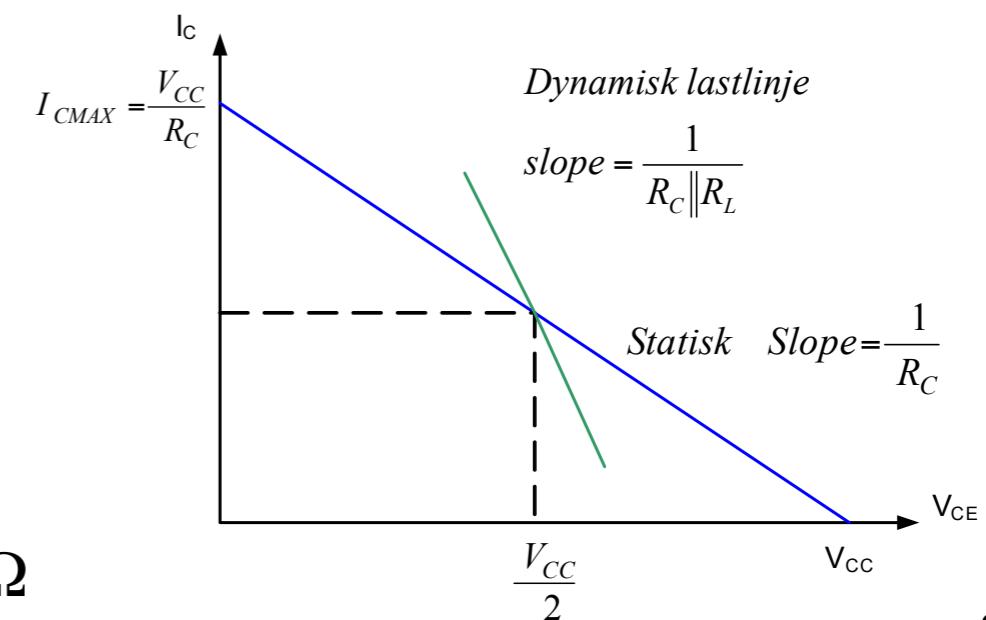
$$R_{inn} = r_\pi + (\beta + 1)R_E = 2,5 \text{ k} + 201 \cdot 1 \text{ k} = 203,5 \text{ k}\Omega$$

Lastlinjen endres i forhold til DC-modellen



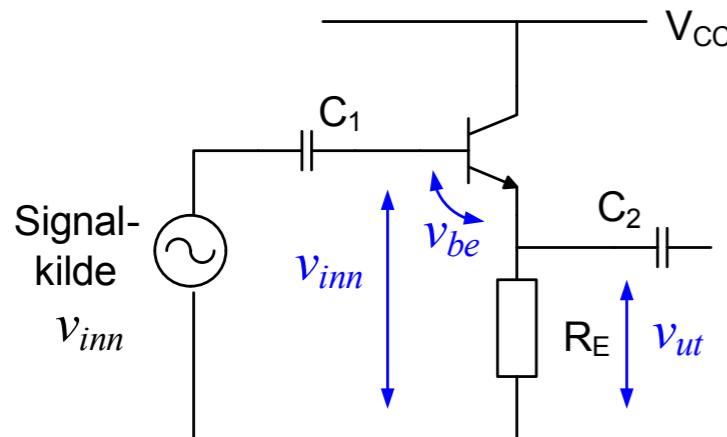
generatoren ser inn mot

$$R_{INN} = R_{B1} \parallel R_{B2} \parallel R_{inn} \approx 10 \text{ k}\Omega$$

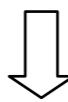


# Transistorforsterker

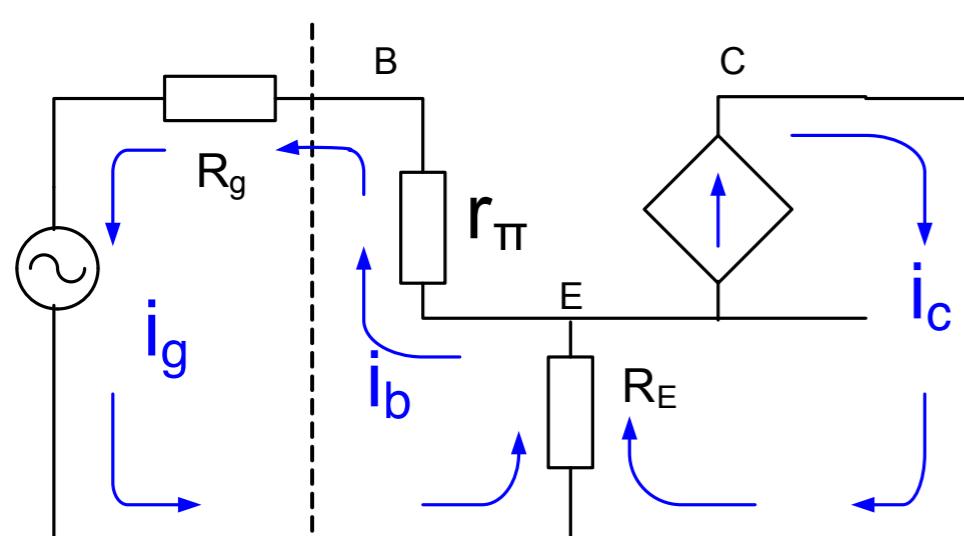
EMITTERFØLGER -Signalet hentes ut over emittermotstanden



Forenklet skjema – utelatt spenningsdeler på basen



Småsignalmodellen



$$\rightarrow i_b = \frac{v_{inn}}{r_\pi + (\beta + 1) \cdot R_E}$$

$$A_v = \frac{v_{ut}}{v_{inn}} = \frac{(\beta + 1) \cdot R_E}{r_\pi + (\beta + 1) \cdot R_E} \cong 1$$

Betrakter signalspenningene  $v_{inn} = v_{be} + v_{ut}$   
 $v_{be}$  er liten  $\rightarrow v_{inn} \approx v_{ut}$

Ser på tilkoplingen av en signalgenerator med indre motstand  $R_g$  og signal  $v_{inn}$

Vi ser bort fra kondensatorene – de stopper DC men slipper signalet uhindret igjennom

$$i_e = (\beta + 1) \cdot i_b \rightarrow i_g = i_b \rightarrow i_e = (\beta + 1) \cdot i_g$$

Strømforsterkning

$$A_i = \frac{i_e}{i_g} = (\beta + 1) \quad \text{Effektforsterkning} \quad P = v \cdot i$$

$$v_{ut} + v_{be} = v_{inn} \rightarrow (\beta + 1) \cdot i_b \cdot R_E + i_b \cdot r_\pi = v_{inn}$$

$$v_{ut} = i_e \cdot R_E = (\beta + 1) \cdot i_b \cdot R_E = \frac{(\beta + 1) \cdot R_E \cdot v_{inn}}{r_\pi + (\beta + 1) \cdot R_E}$$

Stor strømforsterkning  
- Ingen spenningsforsterkning