

State machines and large designs

Lecture #4

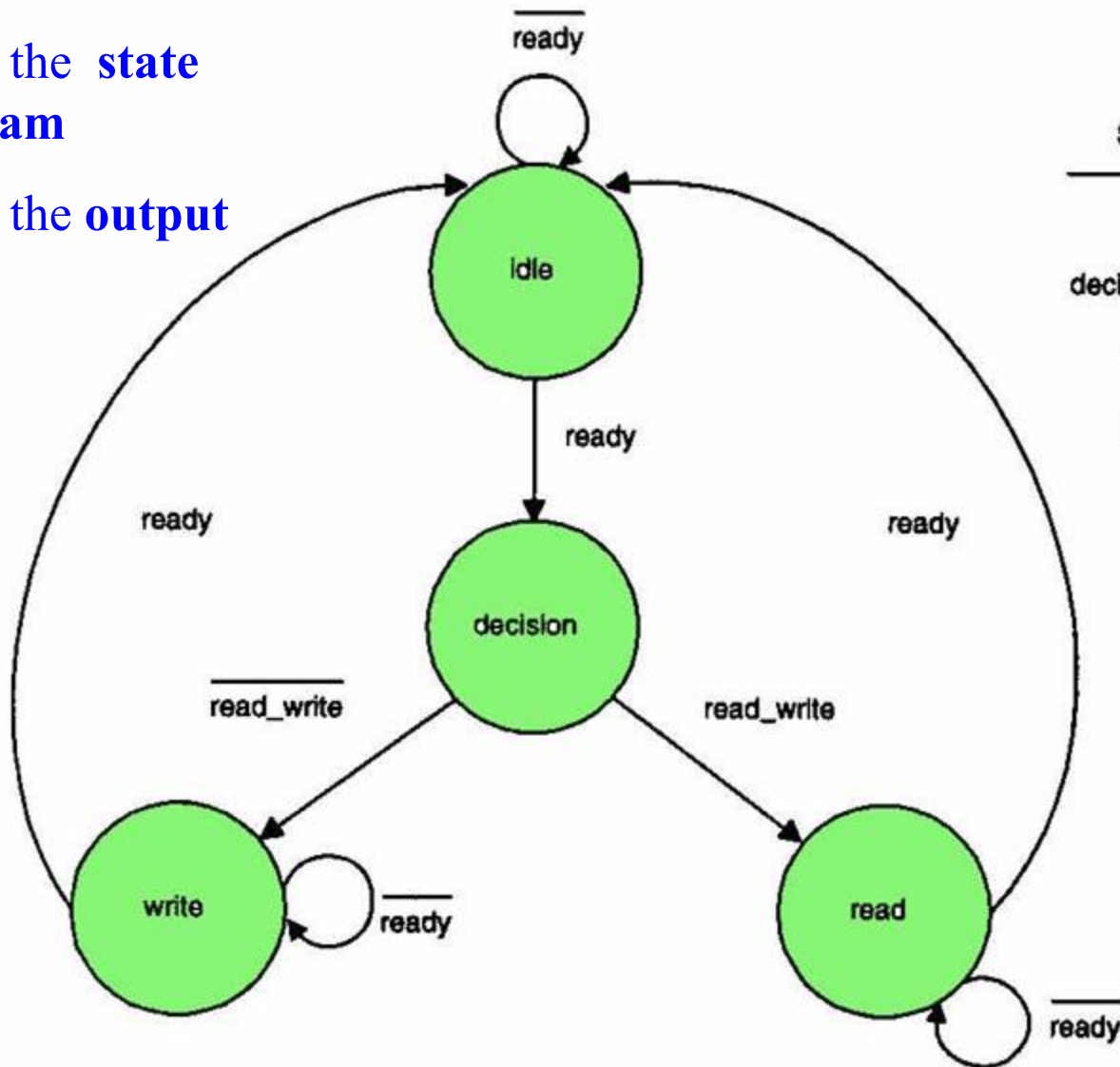


FSMs- Finite State Machines

Traditional method:



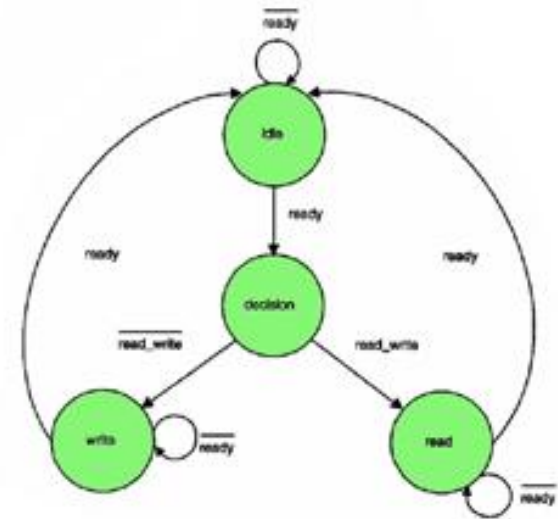
- 1) Make the **state diagram**
- 2) Make the **output table**



| State | Outputs | |
|----------|---------|----|
| | oe | we |
| idle | 0 | 0 |
| decision | 0 | 0 |
| write | 0 | 1 |
| read | 1 | 0 |

Figure 5-1 Simple state machine

3) Make the **state table** from the state diagram and the output table



Present state and input

Next state and output

| PS | | read_write ready | | NS Q_0Q_1 | | Outputs | |
|----------|----------|---------------------|----|-------------|----|---------|----|
| State | q_0q_1 | 00 | 01 | 11 | 10 | oe | we |
| idle | 00 | 00 | 01 | 01 | 00 | 0 | 0 |
| decision | 01 | 11 | 11 | 10 | 10 | 0 | 0 |
| write | 11 | 11 | 00 | 00 | 11 | 0 | 1 |
| read | 10 | 10 | 00 | 00 | 10 | 1 | 0 |

- Write the Boolean equations from the state table
- Use **Karnaugh diagrams** to simplify the Boolean equations

| read_write, ready | | Q_0 | | | |
|----------------------|---|----------|----|----|----|
| | | q_0q_1 | 00 | 01 | 11 |
| 00 | 0 | 0 | 0 | 0 | 0 |
| 01 | 1 | 1 | 1 | 1 | 1 |
| 11 | 1 | 0 | 0 | 1 | 1 |
| 10 | 1 | 0 | 0 | 1 | 1 |

$$Q_0 = \bar{q}_0q_1 + q_0\overline{\text{ready}}$$

| read_write, ready | | Q_1 | | | |
|----------------------|---|----------|----|----|----|
| | | q_0q_1 | 00 | 01 | 11 |
| 00 | 0 | 1 | 1 | 0 | 0 |
| 01 | 1 | 1 | 0 | 0 | 0 |
| 11 | 1 | 0 | 0 | 1 | 1 |
| 10 | 0 | 0 | 0 | 0 | 0 |

$$Q_1 = \bar{q}_0\bar{q}_1\overline{\text{ready}} + \bar{q}_0q_1\overline{\text{read_write}} + q_0q_1\overline{\text{ready}}$$

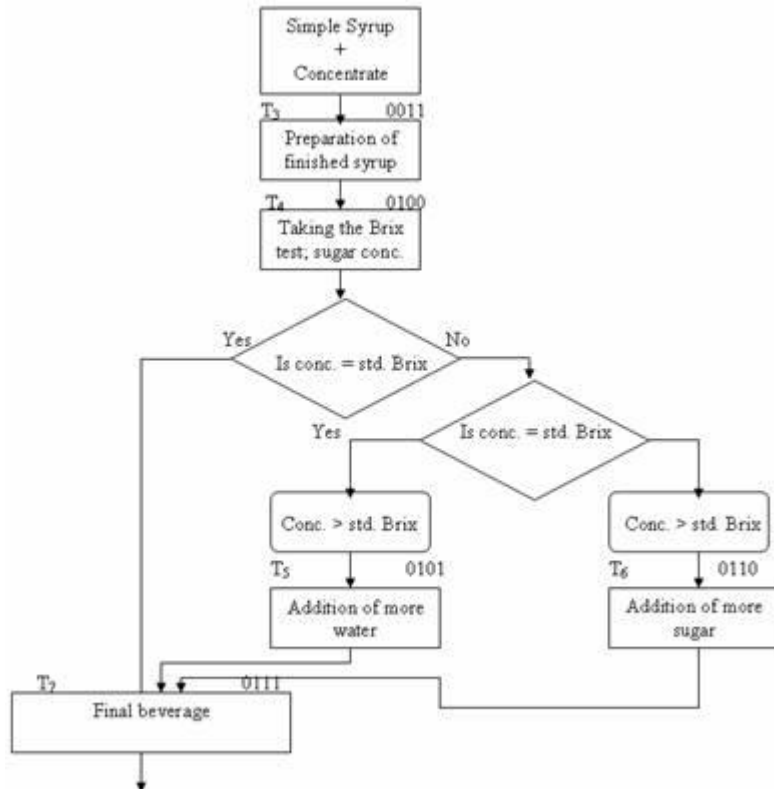
$$\text{oe} = q_0\bar{q}_1$$

$$\text{we} = q_0q_1$$

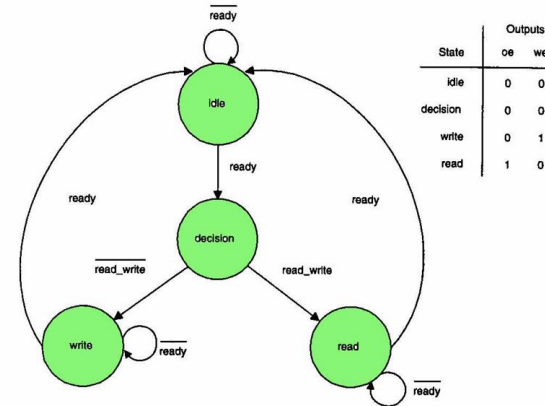
FSM design with VHDL



Algorithmic state machine (ASM) chart
(Flytskjema/Flowchart)

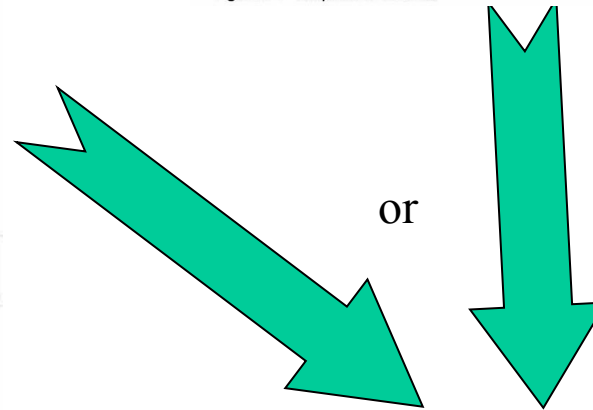


State diagram



| State | Outputs | |
|----------|---------|----|
| | oe | we |
| idle | 0 | 0 |
| decision | 0 | 0 |
| write | 0 | 1 |
| read | 1 | 0 |

Figure 5-1 Simple state machine



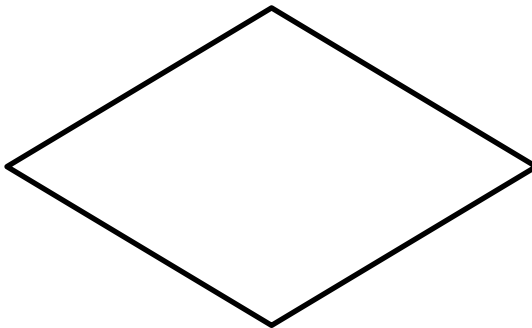
Code the FSM in VHDL

Easy to code the FSM (from the ASM) using VHDL!

Flowchart elements



Tilstandsboks (State box)

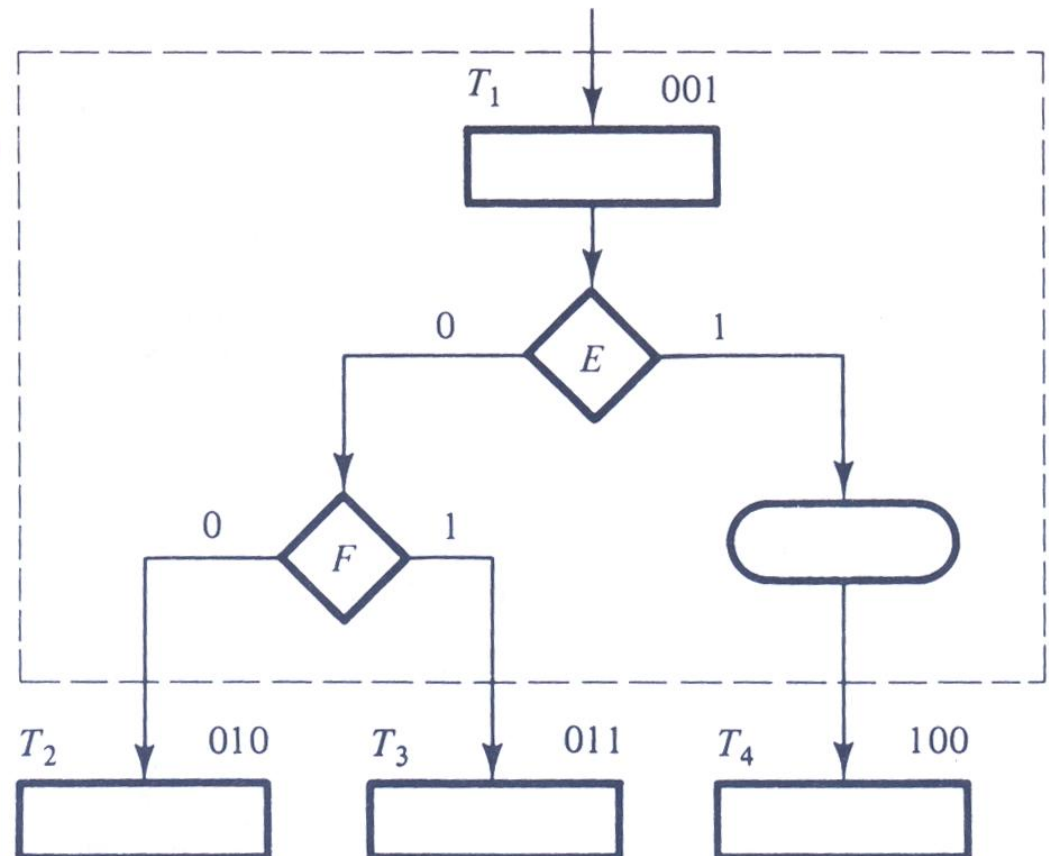
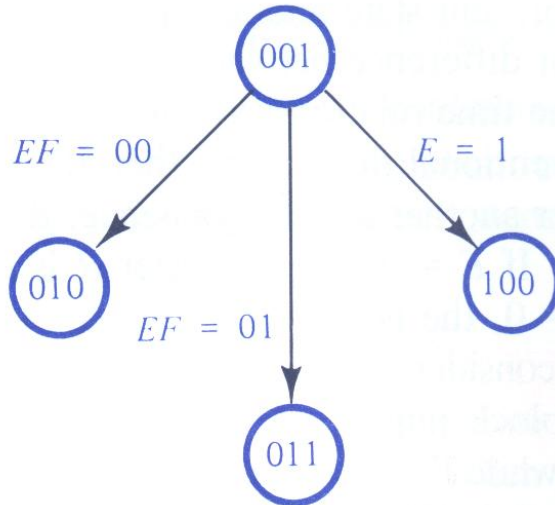


Valgboks (decision box)



Betingelsesboks (Conditional output box)

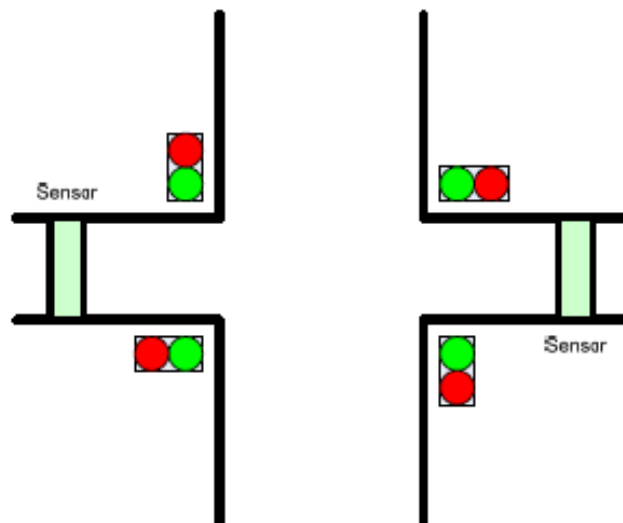
State diagram vs ASM chart



Eksempel – Styring av trafikklys

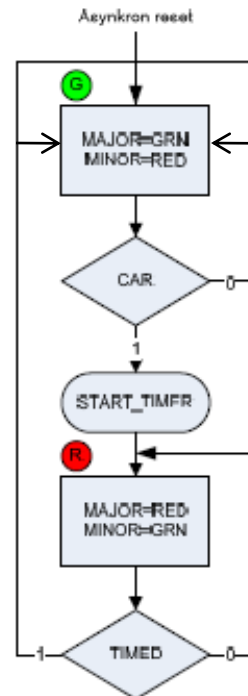
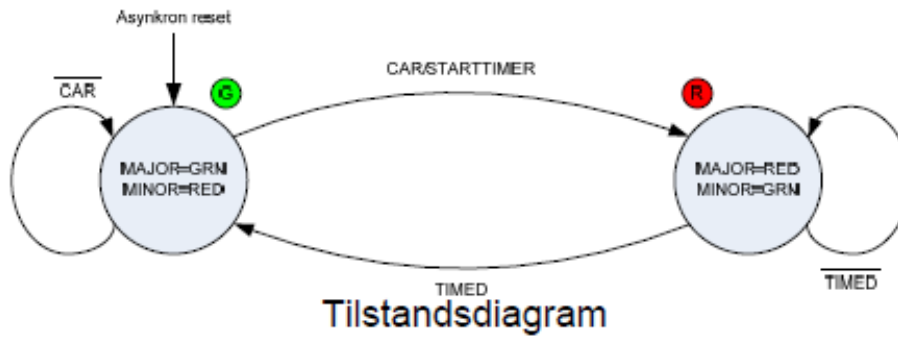


- Eksempel: Enkel trafikklysstyring mellom en hovedvei og to mindre sideveier
- Virkemåte:
 - Sensorer detekterer at det kommer en bil på en sidevei.
 - Gir sideveien grønt lys/hovedvei rødt
 - Sideveien har grønt lys i en viss tid gitt av en timer

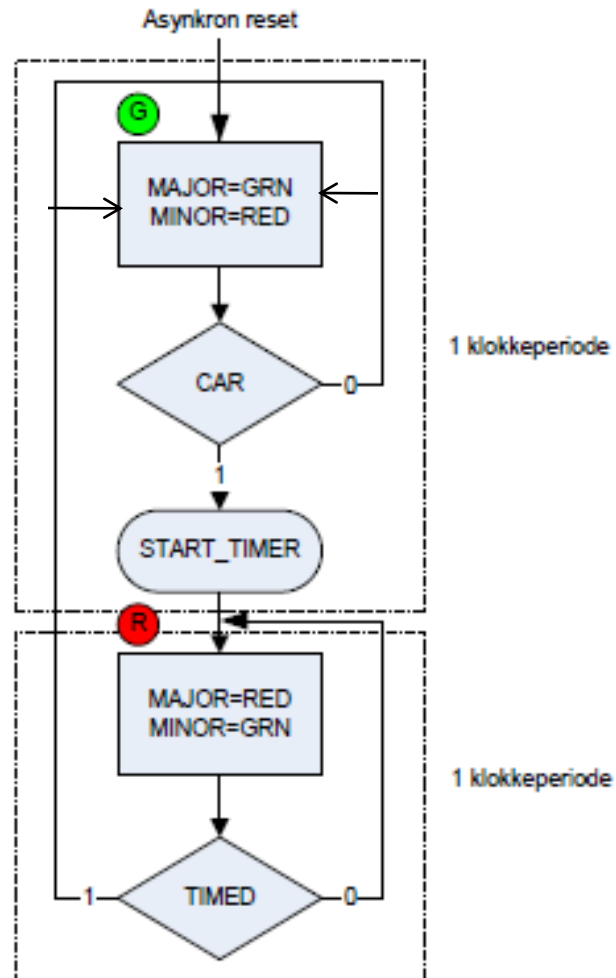


See example in
chapter 5.3

Beskrivelse av tilstandsmaskiner



ASM (Algorithmic State Machine) diagram

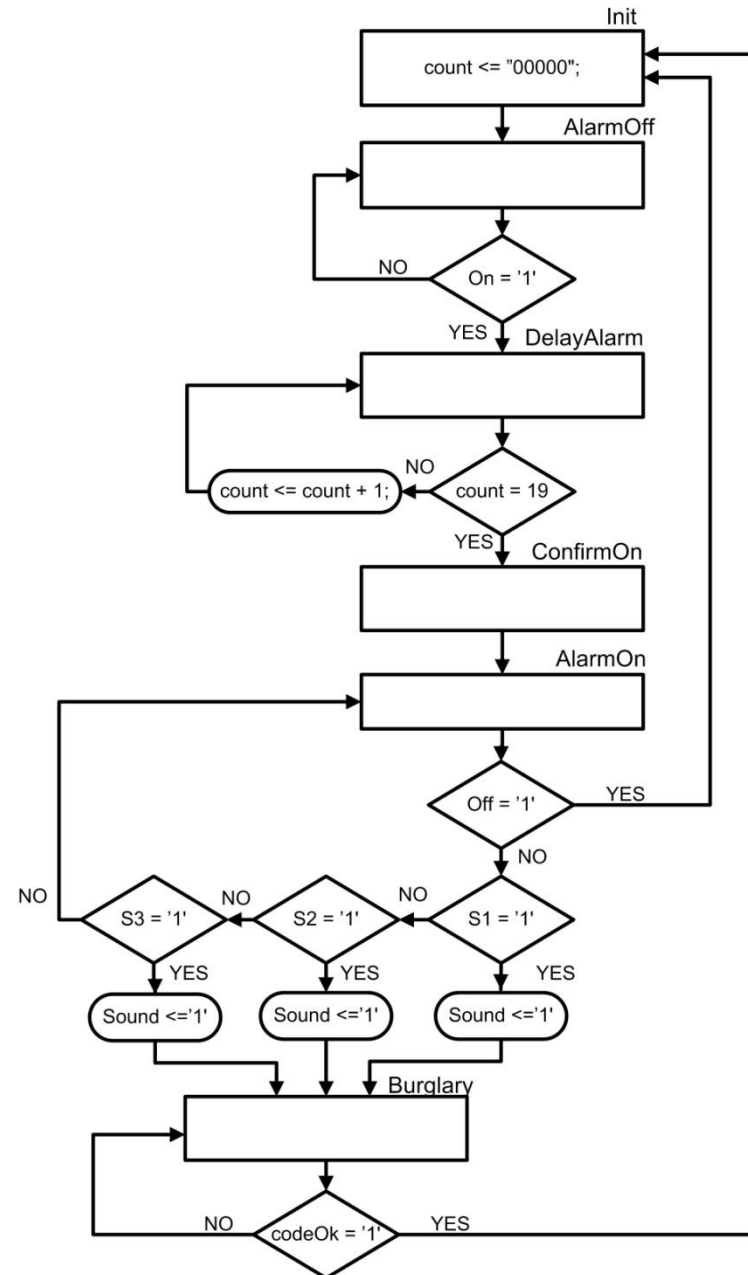


The ASM chart
contains implicit
timing information !

- En tilstand varer fra en tilstandsboks til neste.
- Dette tilsvarer en klokkeperiode

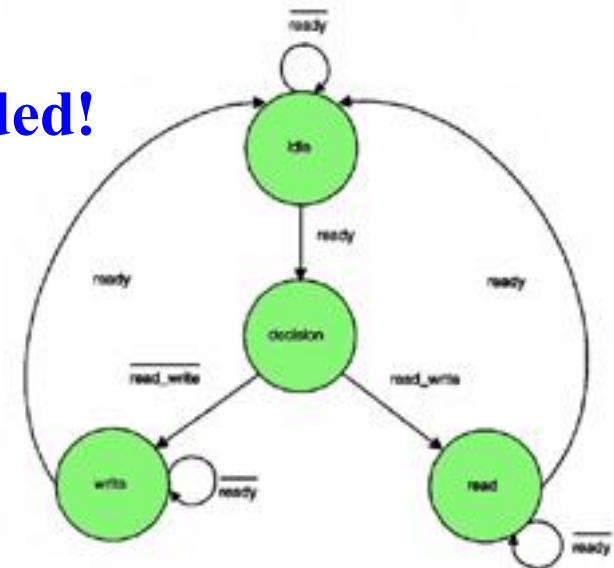
ASM chart guidelines

- Yes/No (1/0) labels together with decision boxes
- Arrows to show the program flow
- State names in top right/left corner of the state boxes
- Selected outputs listed in conditional output boxes and inside the state boxes
- Test condition listed in the decision boxes



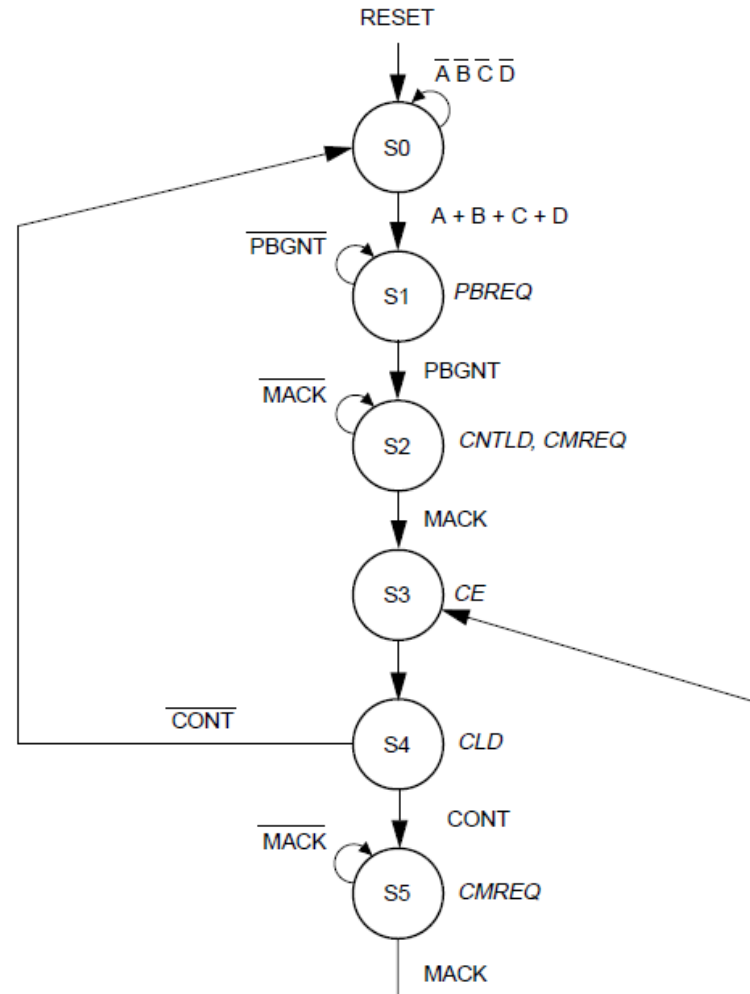
State machines in VHDL

- An ASM chart (or a state diagram) can easily be translated to a VHDL description!
- In VHDL the state machine can be described in two different ways:
 - 1-process FSM ← **Recommended!**
 - 2-process FSM
 - One process describes the combinational logic, and another describes synchronization of state transitions to the clock



Example II

- The state S0 can be realized when state S4 is asserted and the input CONT is low, or it remains at state S0 if all four inputs—A, B, C, and D—are low.
- Similarly, state S1 can be achieved when state S0 is asserted and any one of the four inputs—A, B, C, or D—is high, or it remains at the same state (S1) if the input PBGNT is low



VHDL code

Example II

2-process FSM



```
--library ieee;
--use ieee.std_logic_1164.all;
entity DMASM is
  port (A,B,C,D: in bit;
        PBGNT, MACK, CONT : in bit;
        RST, CLK : in bit;
        PBREQ, CMREQ, CE, CNTLD, CLD : out bit);
end DMASM;

architecture BEHAVE of DMASM is
  type STATE is (S0, S1, S2, S3, S4, S5);
  signal CURRENT_STATE, NEXT_STATE: STATE;
begin
  SEQ: process (RST, CLK)
  begin
    if (RST = '0') then
      CURRENT_STATE <= S0;
    elsif (CLK' event and CLK = '1') then
      CURRENT_STATE <= NEXT_STATE;
    end if;
  end process;

  COMB: process (CURRENT_STATE, A, B, C, D, PBGNT, MACK, CONT)
  begin
    PBREQ <= '0';
    CMREQ <= '0';
    CE <= '0';
    CNTLD <= '0';
    CLD <= '0';
    case CURRENT_STATE is
      when S0 =>
        if (A = '1' or B = '1' or C = '1' or D = '1') then
          NEXT_STATE <= S1;
        else
          NEXT_STATE <= S0;
        end if;
      when S1 => PBREQ <= '1';
        if (PBGNT = '1') then
          NEXT_STATE <= S2;
        else
          NEXT_STATE <= S1;
        end if;
      when S2 => CNTLD <= '1'; CMREQ <= '1';
        if (MACK = '1') then
          NEXT_STATE <= S3;
        else
          NEXT_STATE <= S2;
        end if;
      when S3 => CE <= '1';
        NEXT_STATE <= S4;
      when S4 => CLD <= '1';
        if (CONT = '1') then
          NEXT_STATE <= S5;
        else
          NEXT_STATE <= S0;
        end if;
      when S5 => CMREQ <= '1';
        if (MACK = '1') then
          NEXT_STATE <= S3;
        else
          NEXT_STATE <= S5;
        end if;
    end case;
  end process;
end BEHAVE;
```

Synchronization
to the clock

combinational
logic

Example: 2 process FSM



```
library ieee;
use ieee.std_logic_1164.all;
entity memory_controller is port (
    reset, read_write, ready, burst, clk    : in std_logic;
    bus_id                                  : in std_logic_vector(7 downto 0);
    oe, we                                   : out std_logic;
    addr                                    : out std_logic_vector(1 downto 0));
end memory_controller;

architecture state_machine of memory_controller is
    type StateType is (idle, decision, read1, read2, read3, read4, write);
    signal present_state, next_state : StateType;
begin
    state_comb:process(reset, bus_id, present_state, burst, read_write, ready) begin
        if (reset = '1') then
            oe <= '-'; we <= '-'; addr <= "--";
            next_state <= idle;
        else
            case present_state is
                ...
            end case;
        end if;
    end process state_comb;

    state_clocked:process(clk) begin
        if rising_edge(clk) then
            present_state <= next_state;
        end if;
    end process state_clocked;
end;
```

Combinational
logic

Synchronous
logic



Asynchronous reset in 2-process FSM

```
state_clocked:process(clk, reset) begin
  if reset= '1' then
    present_state <= idle;
  elsif rising_edge(clk) then
    present_state <= next_state;
  end if;
end process state_clocked;
```



1-process FSM

Functionally identical to the 2-process FSM, and the same logic is produced

```
architecture state_machine of memory_controller is
    type StateType is (idle, decision, read1, read2, read3, read4, write);
    signal state : StateType;
begin
state_tr:process(reset, clk) begin
    -- one process fsm
    -- asynchronous reset
    if reset = '1' then
        state <= idle;
    elsif rising_edge(clk) then
        -- synchronization to clk
        -- state transitions defined
        case state is
            when idle =>
                if (bus_id = "11110011") then
                    state <= decision;
                else
                    -- not req'd; for clarity
                    state <= idle;
                end if;
            when decision=>
                if (read_write = '1') then
                    state <= read1;
                else
                    --read_write='0'
                    state <= write;
                end if;
        end case;
    end process;
end architecture;
```



Example: use of state machine for ADC (Analog to Digital Converter) control

[ADC data sheet](#)

[VHDL code](#)

Output from a state machine



1)

```
FSM: process (reset, clk)
begin
  if (reset = '1') then
    state <= idle;

  elsif rising_edge (clk) then
    case state is
      when idle =>
        oe <= '1';
        if (input1 = '1') then
          state <= s1;
        else
          state <= idle;
        end if;
      when s1 =>
        oe <= '0';

        .....
```

2)

Recommended!



```
FSM: process (reset, clk)
begin
  if (reset = '1') then
    state <= idle;

  elsif rising_edge (clk) then
    oe <= '1'; -- Default value
    case state is
      when idle =>
        if (input1 = '1') then
          state <= s1;
        else
          state <= idle;
        end if;
      when s1 =>
        oe <= '0';

        .....
```

3)

```
FSM: process (reset, clk)
begin
  if (reset = '1') then
    state <= idle;

  elsif rising_edge (clk) then
    case state is
      when idle =>
        if (input1 = '1') then
          state <= s1;
        else
          state <= idle;
        end if;
      when s1 =>
        .....
```

end process;

```
with state select
  oe <= '1' when idle,
  '0' when S1;
end state_machine;
```



Example: Use of default values in FSMs

```
FSM_CONF_READ:  
process(clk)  
begin  
    if falling_edge(clk) then  
        .....  
        -- set default values  
        cs          <= '0' ;  
        sr_enable <= '0' ;  
        DataReady <= '0' ;  
        Din        <= '0' ;  
  
        case state is  
            .....  
            when PowerUp =>
```

State encoding and enumeration type



If the number of states is not 2^n there will be undefined states!

```
type StateType is (idle, decision, read1, read2, read3,  
                  read4, write);  
signal state : StateType;
```

Table 5-8 Sequential encoding for an enumeration type

| State | q_0 | q_1 | q_2 |
|-----------|-------|-------|-------|
| idle | 0 | 0 | 0 |
| decision | 0 | 0 | 1 |
| read1 | 0 | 1 | 0 |
| read2 | 0 | 1 | 1 |
| read3 | 1 | 0 | 0 |
| read4 | 1 | 0 | 1 |
| write | 1 | 1 | 0 |
| undefined | 1 | 1 | 1 |

Seven states

← Added state; $2^3 = 8$

FSM fault tolerance

- If the state machine gets into an **undefined (illegal) state** it will be **unpredictable** (can e.g. stay in the illegal state forever)
- In safety critical systems it is absolutely necessary to always get to known states
- Demands additional logic to get out of illegal states

```
type StateType is (idel, decision, read1, read2, read3, read4,  
                  write, undefined);
```

Include a state name in the enumerated type for each undefined state

```
case present_state is
```

```
...
```

```
when undefined => next_state <= idle;
```

```
end case;
```

Specify the state transition to get out of illegal states



Example with 3 undefined states

```
type states is (s0, s1, s2, s3, s4, u1, u2, u3);  
signal next_state, present_state: states;  
...  
case present_state is  
    ...  
    when others => next_state <= s0;  
end case;
```




One-Hot encoding of FSM

- Uses one bit for each state
- A state machine with n states needs n flip-flops to store the states
- Only one flip-flop (bit) assigned to '1', indicating present state
- Output logic and next state logic is very simple
- Used in FPGAs (due to many flip-flops)



| State | Sequential | One-Hot |
|---------|------------|----------------------|
| state0 | 00000 | 00000000000000000001 |
| state1 | 00001 | 00000000000000000010 |
| state2 | 00010 | 00000000000000000100 |
| state3 | 00011 | 00000000000000001000 |
| state4 | 00100 | 0000000000000010000 |
| state5 | 00101 | 000000000000100000 |
| state6 | 00110 | 00000000001000000 |
| state7 | 00111 | 00000000010000000 |
| state8 | 01000 | 00000000100000000 |
| state9 | 01001 | 00000001000000000 |
| state10 | 01010 | 00000010000000000 |
| state11 | 01011 | 00000100000000000 |
| state12 | 01100 | 00001000000000000 |
| state13 | 01101 | 00010000000000000 |
| state14 | 01110 | 00100000000000000 |
| state15 | 01111 | 00100000000000000 |
| state16 | 10000 | 01000000000000000 |
| state17 | 10001 | 10000000000000000 |

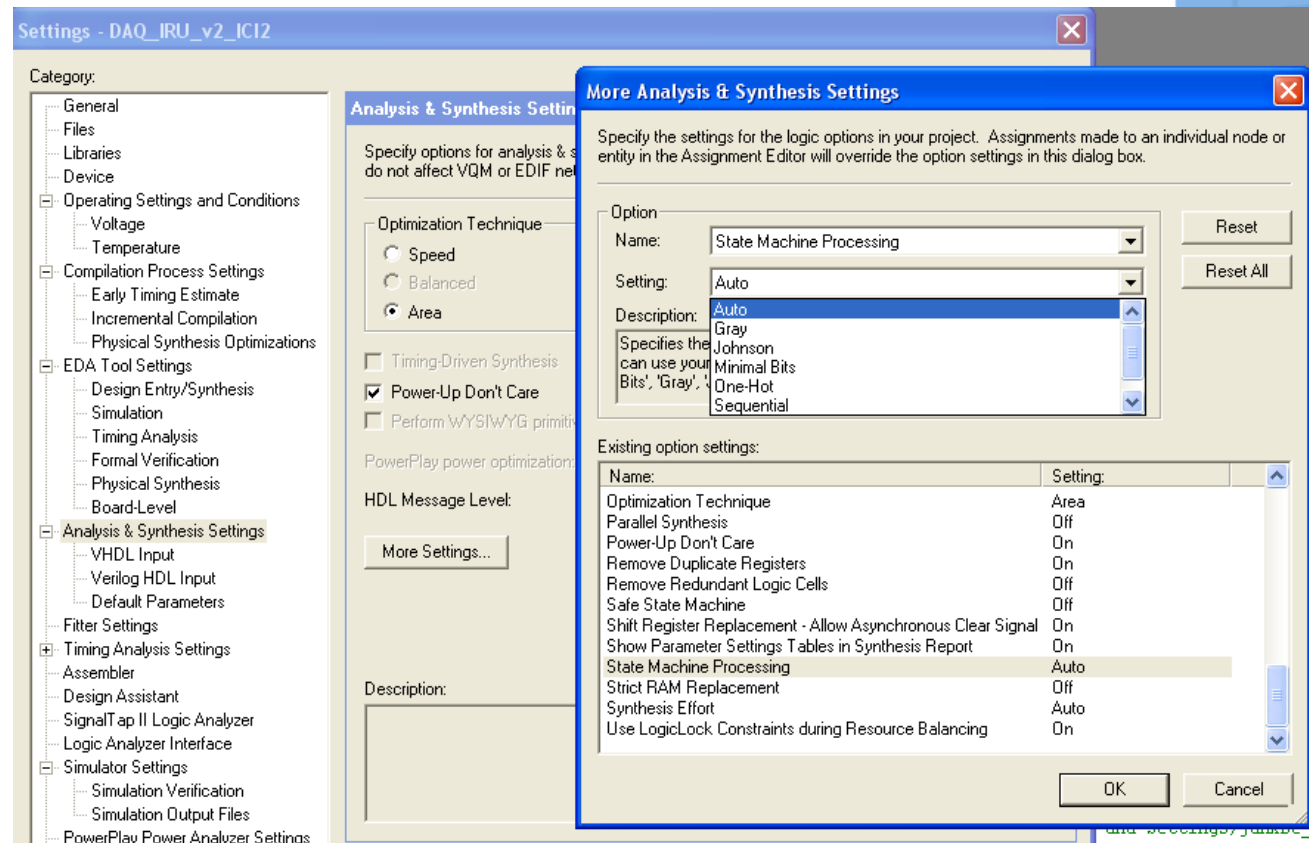
State machine with 18 states

5

18

One-Hot encoding setup in Quartus II

- No change in the VHDL code
- Setup in Quartus II before synthesis (**Assignments – Settings Analysis & Synthesis Settings – More Settings**)



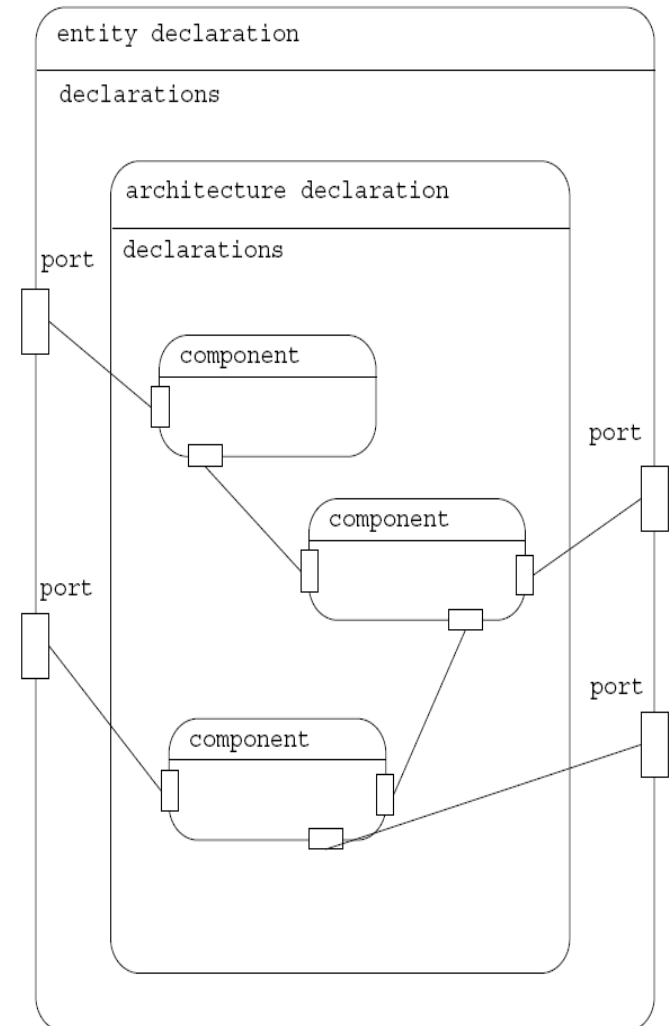
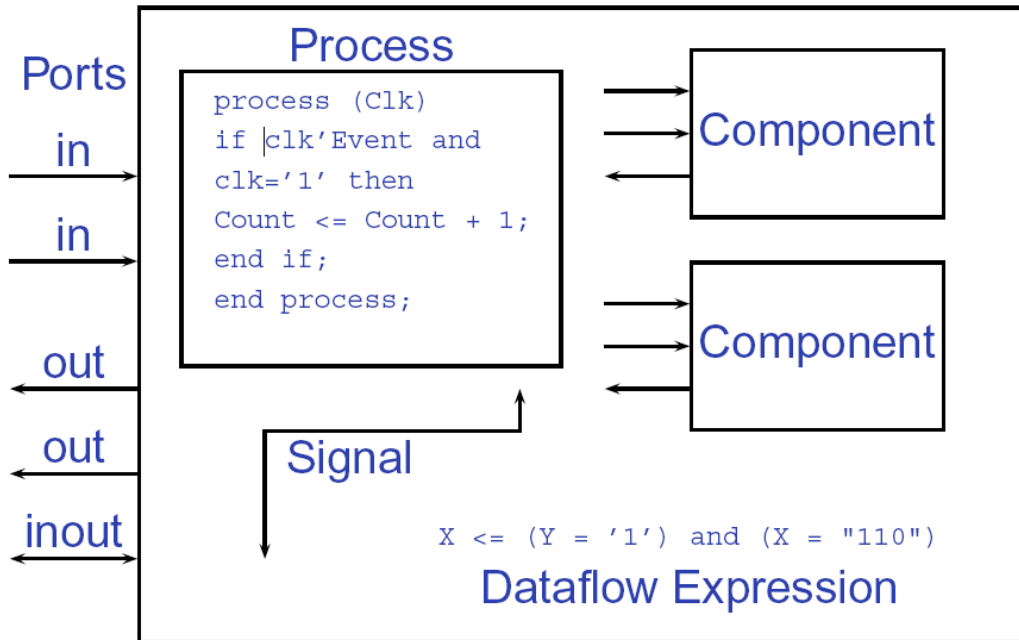
The screenshot shows the 'Settings - DAQ_IRU_v2_IC12' dialog box in Quartus II. The 'Analysis & Synthesis Settings' tab is active, and the 'More Analysis & Synthesis Settings' sub-dialog is open. The 'More Analysis & Synthesis Settings' dialog allows for configuring logic options for the project. The 'Option' section shows 'State Machine Processing' selected, with the 'Setting' set to 'Auto'. The 'Description' field is highlighted, showing the text: 'Specifies the logic options that can use your logic resources. The options are Johnson, Minimal Bits, One-Hot, and Sequential.' The 'Existing option settings' table lists various options and their current settings.

| Name: | Setting: |
|--|----------|
| Optimization Technique | Area |
| Parallel Synthesis | Off |
| Power-Up Don't Care | On |
| Remove Duplicate Registers | On |
| Remove Redundant Logic Cells | Off |
| Safe State Machine | Off |
| Shift Register Replacement - Allow Asynchronous Clear Signal | On |
| Show Parameter Settings Tables in Synthesis Report | On |
| State Machine Processing | Auto |
| Strict RAM Replacement | Off |
| Synthesis Effort | Auto |
| Use LogicLock Constraints during Resource Balancing | On |



”Large designs - Hierarchy”

Block diagrams



Hvordan gå frem for å skrive VHDL kode



■ Tegn et forenklet blokkskjema

- ytre grensesnittet, i form av innsignaler og utsignaler, tilsvarer *entity-deklarasjonen*
- indre strukturen tilsvarer VHDL-beskrivelsens *architecture-del*.

■ Det er ofte hensiktsmessig å dele opp den indre strukturen i en *dataveistruktur* og en *kontrollstruktur*.

■ Dataveistrukturen inneholder elementer som registre, addere, multipleksere, som er sammenkoblet med databusser. Denne strukturen er velegnet til å beskrives i blokkskjema. ***Hvis strukturen er kompleks, lønner det seg å dele den opp i mindre blokker***

■ Kontrollogikkens innsignaler og utsignaler kan beskrives i blokkskjema sammen med dataveistrukturen. Den indre oppbyggingen beskrives best i form av: **sannhetsverditabeller, boolske ligninger og ASM flyt-skjema/tilstandsdiagrammer**

■ Deretter kodes den strukturerte løsningen i VHDL

■ Med utgangspunkt i en veldokumentert struktur skal det normalt være en grei jobb å lage en fungerende kode.

Example of a block diagram

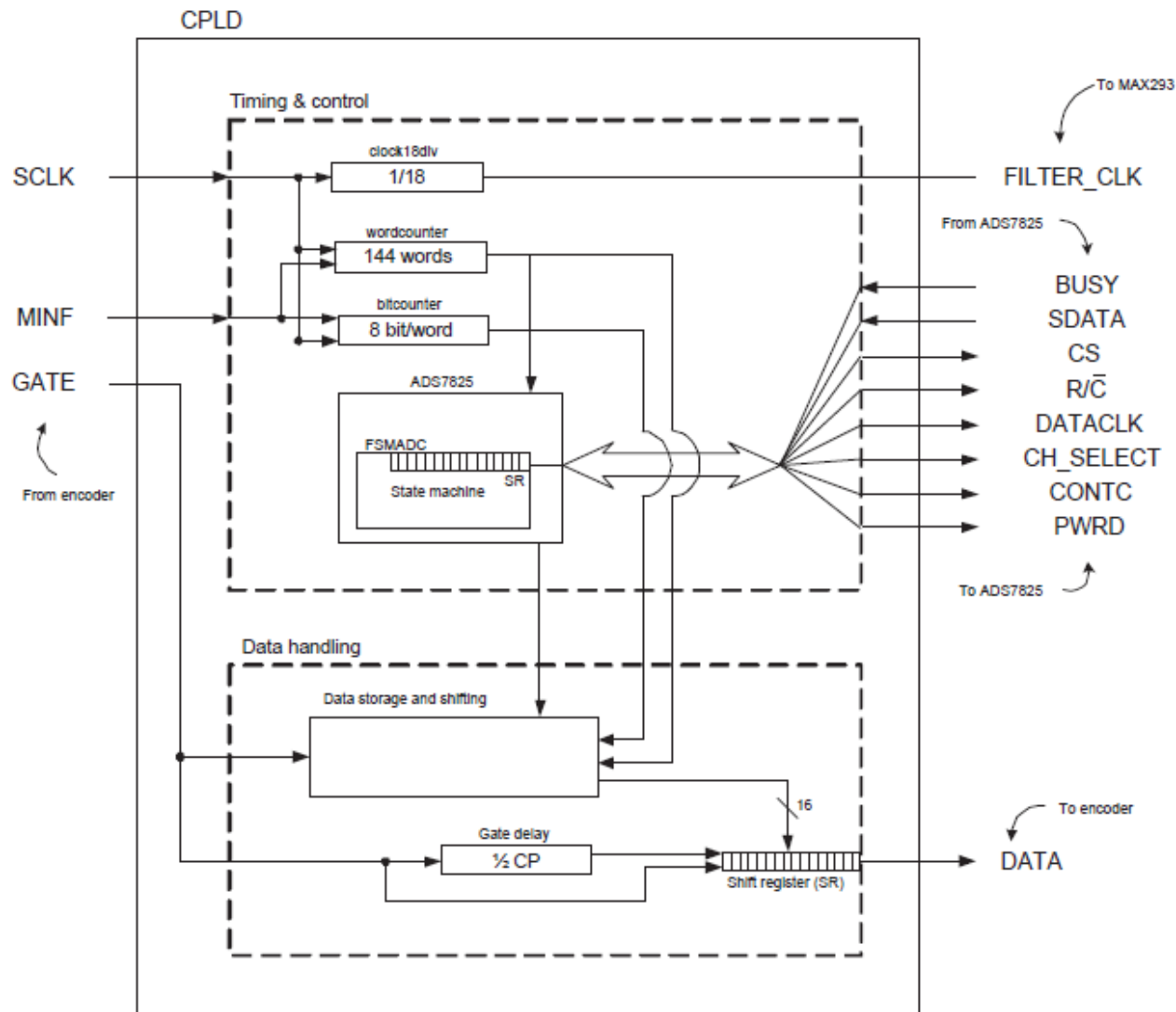


Figure 5.18: Block diagram for the digital logic



Building blocks

■ Library

- library IEEE
- use IEEE.std_logic_1164.all;

} To make the library visible

■ Packages

- use work.my_package.all; -- your own package
- **work** is the directory where the design files are located

■ Components

- An entity used in another entity
- Needs a component declaration to make the component visible
- Needs a component instantiation to connect the component to the top entity – using *port map()*

Component declarations in a package

- A component declaration can be placed in a package
- To use the component in the design the package must be included in the design
- The package must have its own “**library and use**”
- The package is usually in a separate file, (e.g. *my_package.vhd*), and the components are in separate files as well (e.g. *dflop.vhd*)

Without using a package:

```
architecture arch_test_dff of test_dff is
-- Component declaration
component dflop
port(
  d, clk : in  std_logic
  q      :out std_logic);
end component;

-- Declaration of internal signals
signal temp : std_logic;

begin
```

Package example



```
library ieee;
use ieee.std_logic_1164.all;
use work.numeric_std.all;
package regs_pkg is
  component rdff1 port (
    clk, reset: in std_logic;
    d:          in std_logic;
    q:          buffer std_logic);
  end component;
  component rdff
    generic (size: integer := 2);
    port (clk, reset: in std_logic;
          d:          in std_logic_vector(size-1 downto 0);
          q:          buffer std_logic_vector(size-1 downto 0));
  end component;
  component rreg1 port(
    clk, reset, load:in std_logic;
    d:          in std_logic;
    q:          buffer std_logic);
  end component;
end regs_pkg;
```

A larger design using a package



```
library ieee;
use ieee.std_logic_1164.all;
use work.numeric_std.all;
use work.am2901_comps.all; ← Component declaration
entity am2901 is port(
  clk, rst:    in std_logic;
  a, b:       in unsigned(3 downto 0);    -- address inputs
  d:         in unsigned(3 downto 0);    -- direct data
  i:         in std_logic_vector(8 downto 0);
                                     -- micro instruction
  c_n:       in std_logic;              -- carry in
  oe:       in std_logic;              -- output enable
  ram0, ram3: inout std_logic;         -- shift lines to ram
  qs0, qs3: inout std_logic;         -- shift lines to q
  y:        buffer unsigned(3 downto 0); -- data outputs (3-state)
  g_bar,p_bar: buffer std_logic;      -- carry generate, propagate
  ovr:      buffer std_logic;         -- overflow
  c_n4:     buffer std_logic;         -- carry out
  f_0:     buffer std_logic;         -- f = 0
  f3:     buffer std_logic;         -- f(3) w/o 3-state
end am2901;
architecture am2901 of am2901 is
  alias dest_ctl: std_logic_vector(2 downto 0) is i(8 downto 6);
  alias alu_ctl: std_logic_vector(2 downto 0) is i(5 downto 3);
  alias src_ctl: std_logic_vector(2 downto 0) is i(2 downto 0);
  signal ad, bd: unsigned(3 downto 0);
  signal q: unsigned(3 downto 0);
  signal r, s: unsigned(3 downto 0);
  signal f: unsigned(3 downto 0);
begin
  -- instantiate and connect components
  u1: ram_regs port map(clk => clk, rst => rst, a => a, b => b, f => f,
    dest_ctl => dest_ctl, ram0 => ram0, ram3 => ram3,
    ad => ad, bd => bd);
  u2: q_reg port map(clk => clk, rst => rst, f => f, dest_ctl => dest_ctl,
    qs0 => qs0, qs3 => qs3, q => q);
  u3: src_op port map(d => d, ad => ad, bd => bd, q => q,
    src_ctl => src_ctl, r => r, s => s);
  u4: alu port map(r => r, s => s, c_n => c_n, alu_ctl => alu_ctl,
    f => f, g_bar => g_bar, p_bar => p_bar,
    c_n4 => c_n4, ovr => ovr);
  u5: out_mux port map(ad => ad, f => f, dest_ctl => dest_ctl,
    oe => oe, y => y);
  -- define f_0 and f3 outputs
  f_0 <= '0' when f = "0000" else 'Z';
  f3 <= f(3);
end am2901;
```

← Component instantiation
(Komponenttilordninger)

Generics



- The width of a signal can be specified using a parameter
- Useful for **registers** and **counters** with different number of bits; only necessary to make one component

```
-----  
library ieee;  
use ieee.std_logic_1164.all;  
entity rdff is  
  generic (size: integer := 2);  
  port ( clk, reset: in std_logic;  
        d:          in std_logic_vector(size-1 downto 0);  
        q:          buffer std_logic_vector(size-1 downto 0));  
end rdff;
```

Default value

```
architecture archrdff of rdff is  
begin  
  p1: process (reset, clk) begin  
    if reset = '1' then  
      q <= (others => '0');  
    elsif (clk'event and clk='1') then  
      q <= d;  
    end if;  
  end process;  
end archrdff;
```



Generic counter

```
library ieee;
use ieee.std_logic_1164.all;
use work.std_arith.all;
entity ascount is
    generic (CounterSize: integer := 2);
    port(clk, areset, sreset, enable: in std_logic;
         count: buffer std_logic_vector(counterSize-1 downto 0));
end ascount;
architecture archascount of ascount is
begin
    p1: process (areset, clk) begin
        if areset = '1' then
            count <= (others => '0');
        elsif (clk'event and clk='1') then
            if sreset='1' then
                count <= (others => '0');
            elsif enable = '1' then
                count <= count + 1;
            else
                count <= count;
            end if;
        end if;
    end process;
end archascount;
```

Using Generic components

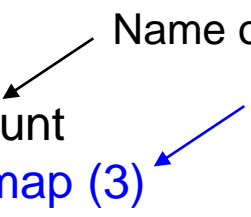
- **generic map** is used to assign a value to a generic component
- If the **generic map** is not specified the default value which is specified in the component is used

Example:

U1: ascount
generic map (3)
port map (clk, areset, sreset, enable, count);

Name of the component

Set the value of the CounterSize parameter
-- or generic map (CounterSize => 3)



Generics Example



- The width of a signal can be specified using a parameter
- Useful for **registers** and **counters** with different number of bits; only necessary to make one component

Entity using a generic

```
entity SR_SerIn_redge is
  generic (
    width : integer := 16);

  port
  (
    clk      : in  std_logic;
    DataIn   : in  std_logic;
    shift_en : in  std_logic;
    DataOut  : out std_logic_vector(width-1 downto 0)
  );
end SR_SerIn_redge;
```

How to use this component with a generic

```
component SR_SerIn_redge
  generic (width : integer);

  port
  (
    clk      : in  std_logic;
    DataIn   : in  std_logic;
    shift_en : in  std_logic;
    DataOut  : out std_logic_vector(width-1 downto 0)
  );
end component;

SR_DATA: SR_SerIn_redge
  generic map (
    width => 12)

  port map (sclk, Dout, sr_enable, Pdata );
```